

INTERFERENCE CHANNELS

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Based on joint work with Chandra Nair (CUHK)

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IEEE Benelux Information Theory Chapter

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Eindhoven Univ. of Technology

Dedication: In honor of Thomas Cover



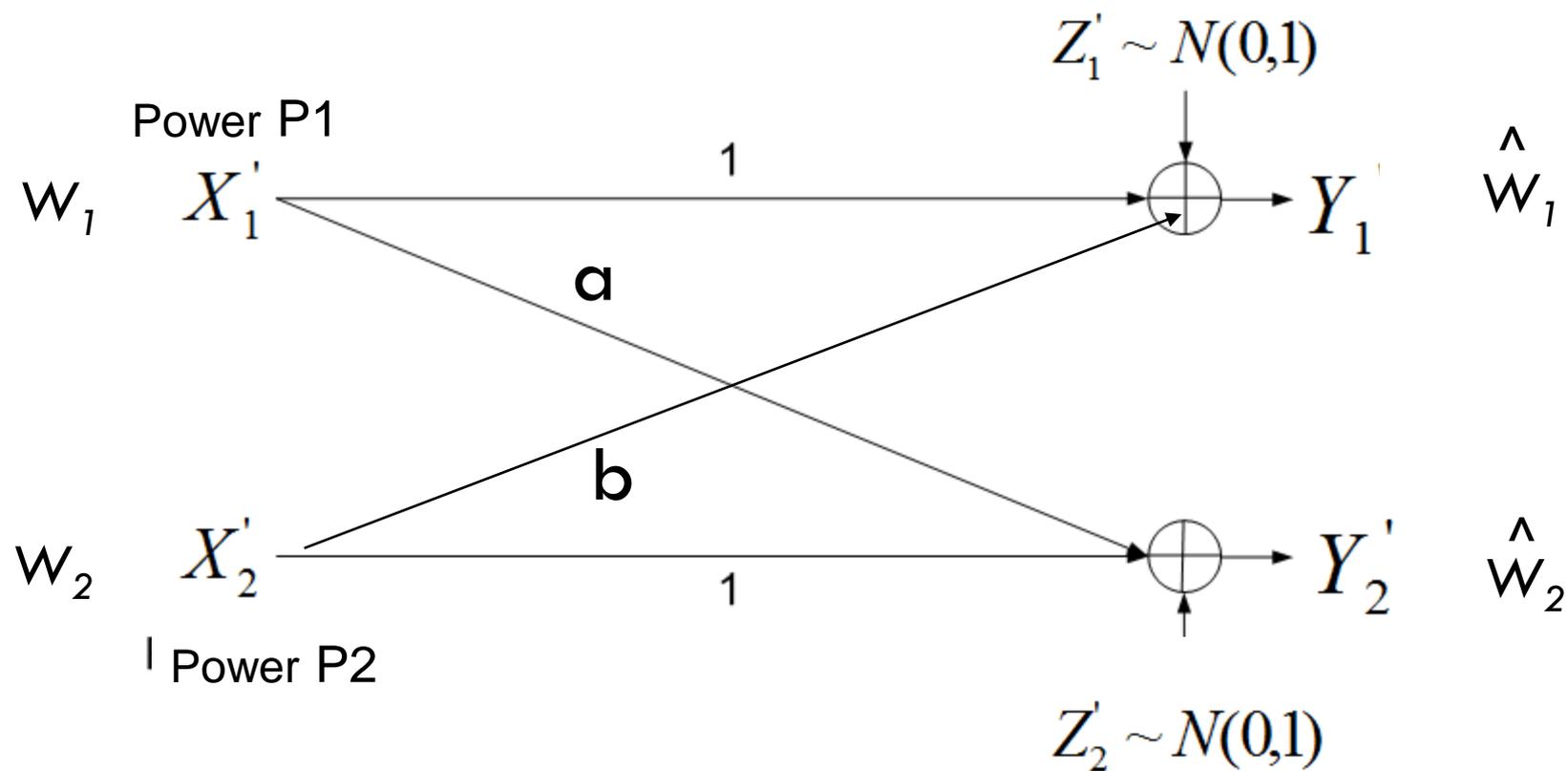
Thomas Cover – 1938-2012

“There are a lot of simple and shocking statements that come out of probability and statistics. It is the existence of these as yet unfound statements that drives my interest in the field.

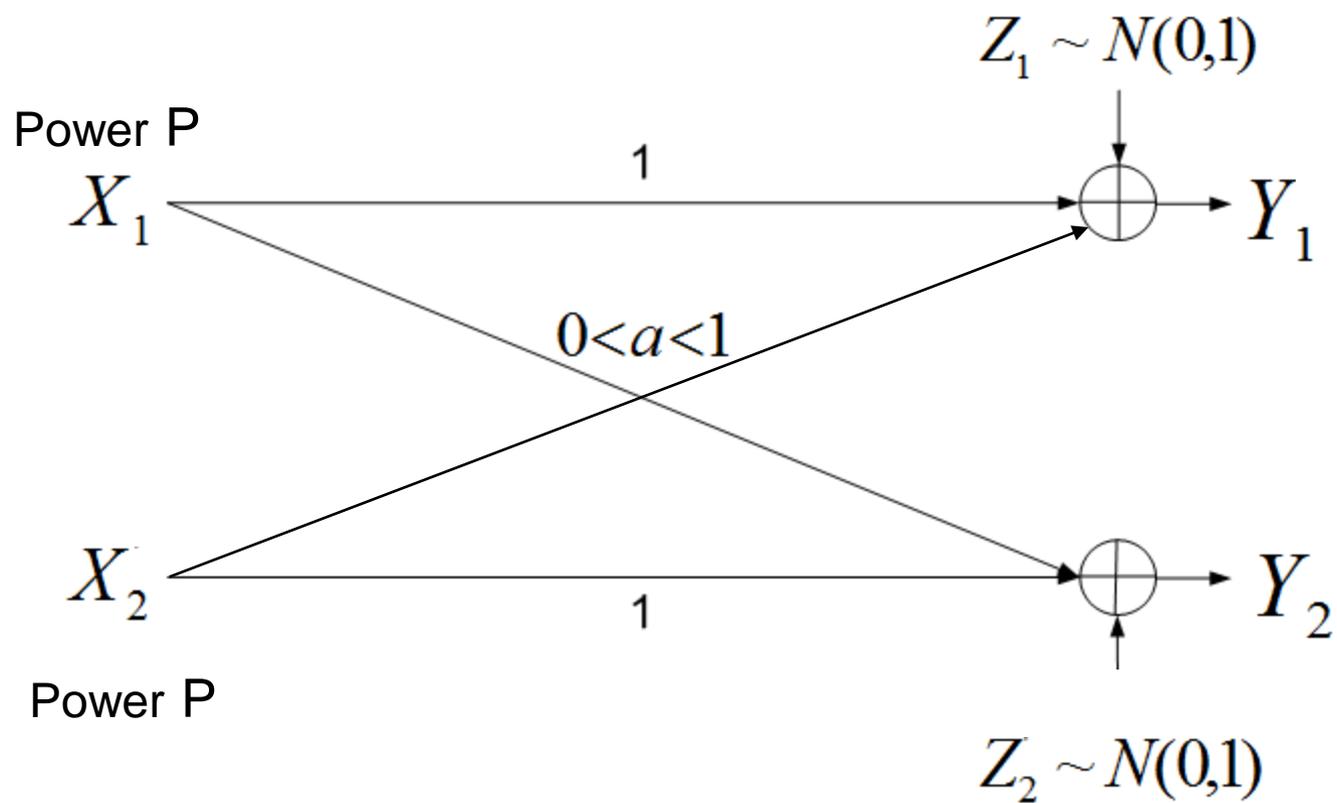
Summary

- Gaussian Interference Channel - standard form
- Brief history
- Z-Interference channel
- Symmetric Interference channel

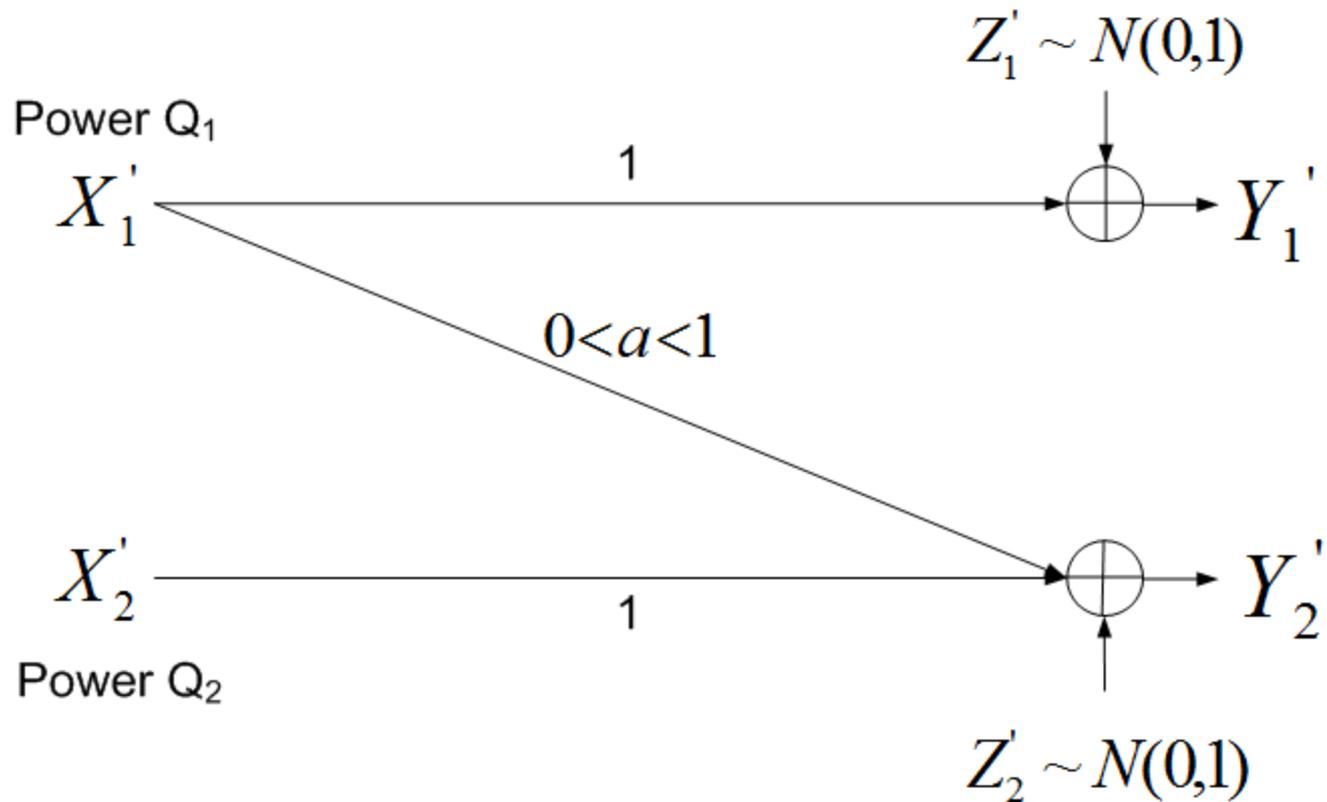
Standard Gaussian Interference Channel



Symmetric Gaussian Interference Channel



Z-Gaussian Interference Channel



The possibilities:

Things that we can do with interference:

1. Ignore (take interference as noise (IAN))
2. Avoid (divide the signal space (TDM/FDM))
3. Partially decode both interfering signals
4. Partially decode one, fully decode the other
5. Fully decode both (only good for strong interference, $a \geq 1$)

Brief history

- Carleial (1975): Very strong interference does not reduce capacity ($a^2 \geq 1+P$)
- Sato (1981), Han and Kobayashi (1981): Strong interference ($a^2 \geq 1$): IFC behaves like 2 MACs
- Motahari, Khandani (2009), Shang, Kramer and Chen (2009), Annapureddy, Veeravalli (2009):
Very weak interference ($2a(1+a^2P) \leq 1$):
 \implies Treat interference as noise (IAN)

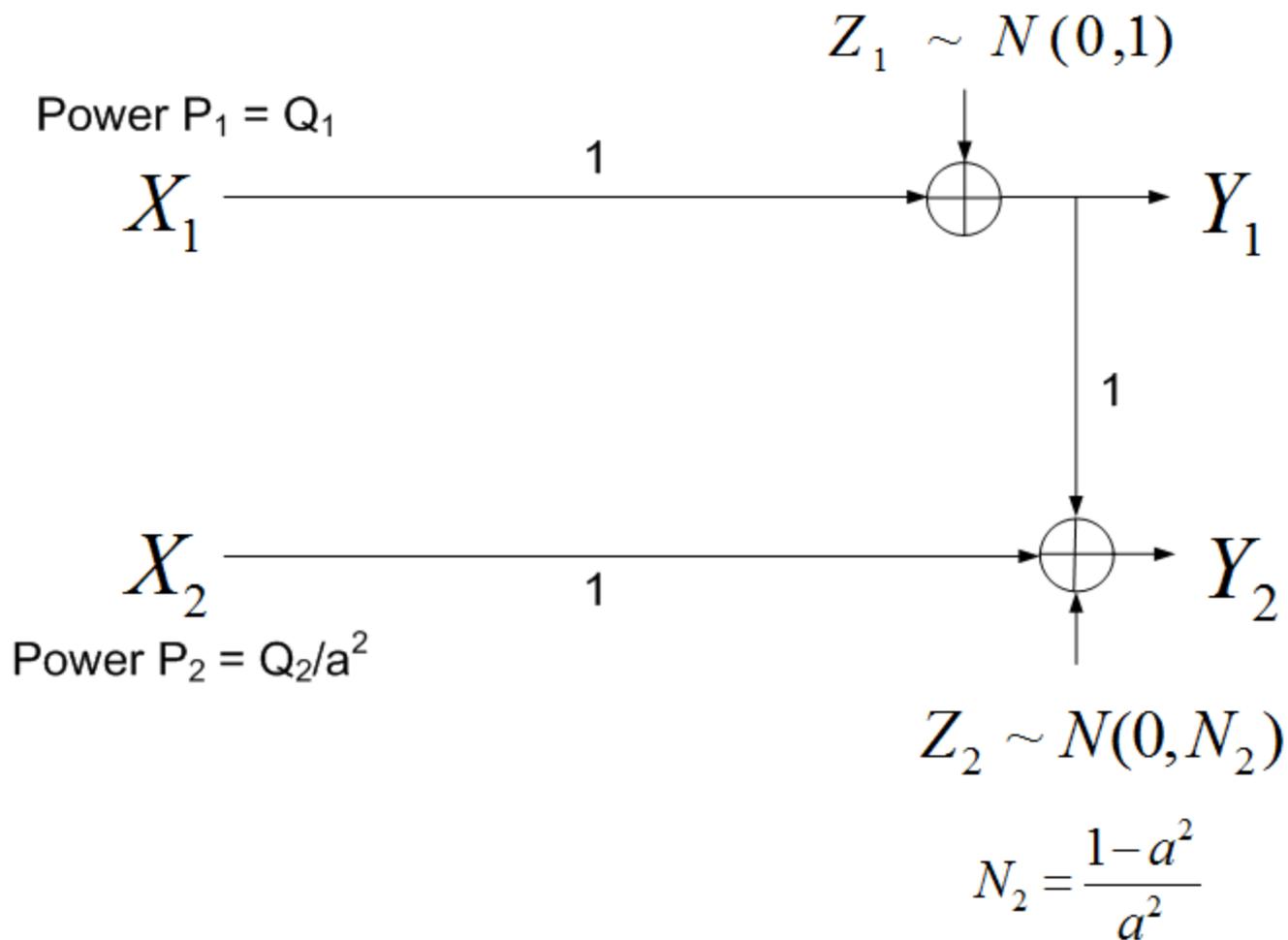
History (continued)

- Sason (2004): Symmetrical superposition to beat TDM – found part of optimal choice for α
- Etkin, Tse, Wang (2008): capacity to within 1 bit, good heuristical choice of $\alpha P = 1/a^2$

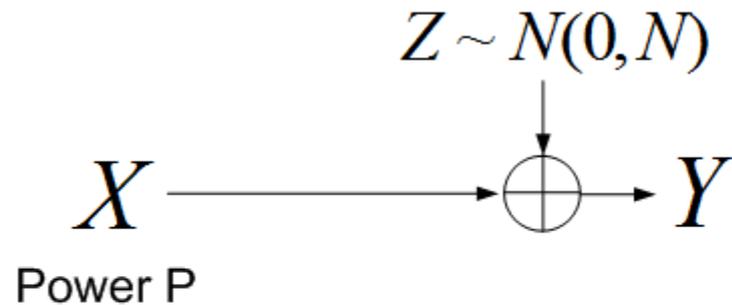
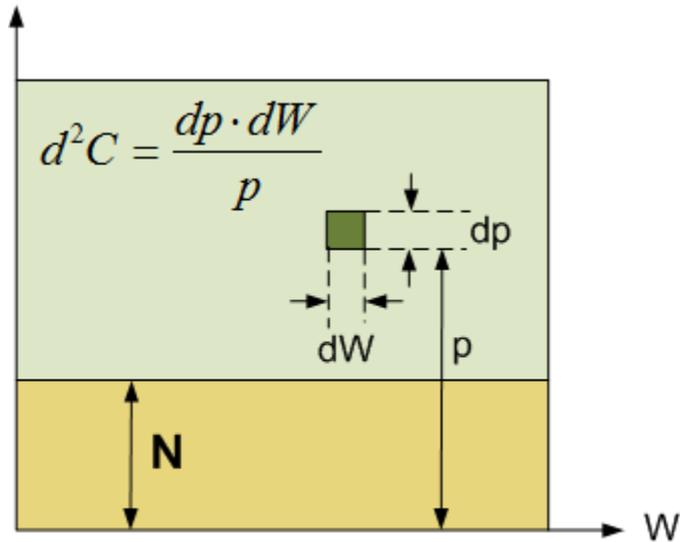
Summary: Z interference Channels

- ❑ Z-Gaussian Interference Channel as a degraded interference channel
- ❑ Discrete Memoryless Channel as a band limited channel
- ❑ Multiplex Region: growing **Noisebergs**
- ❑ Overflow Region: back to superposition

Degraded Gaussian Interference Channel



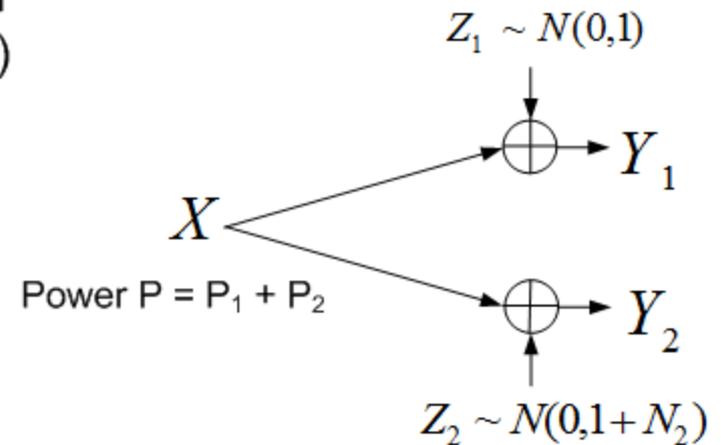
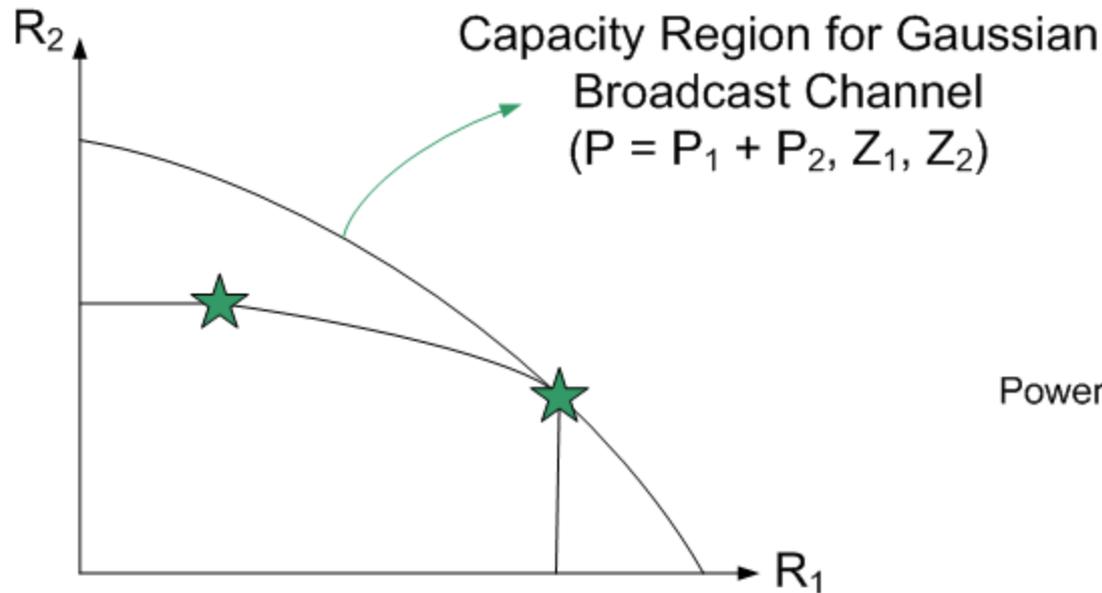
Differential capacity



$$C = \iint d^2C = \frac{1}{2} \log \left(1 + \frac{P}{N} \right)$$

Discrete memoryless channel as a band limited channel

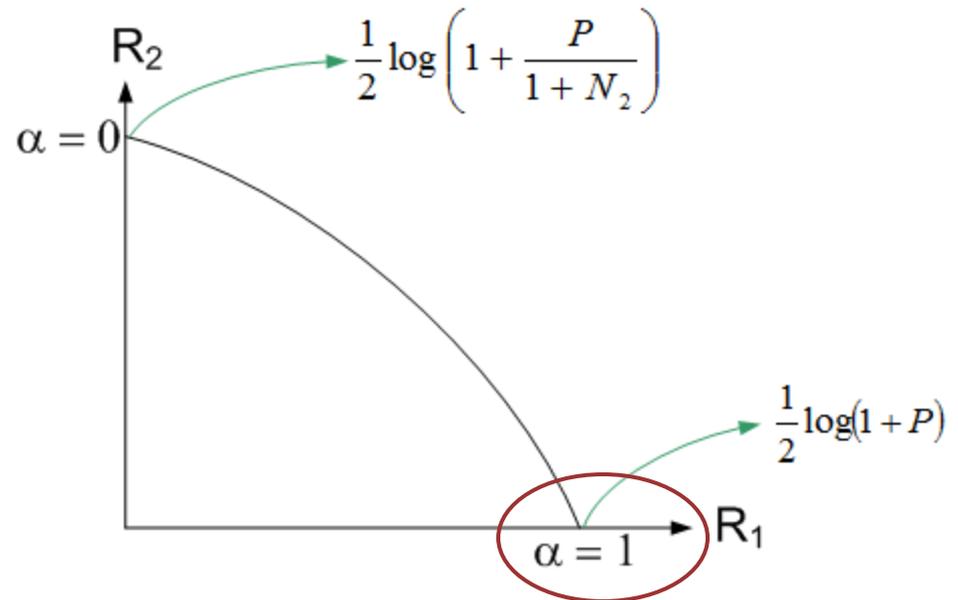
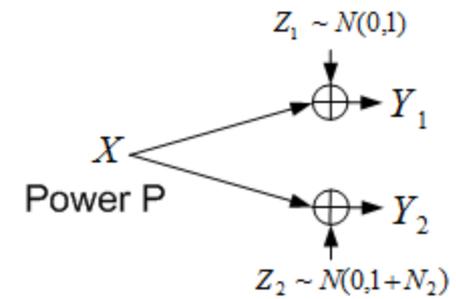
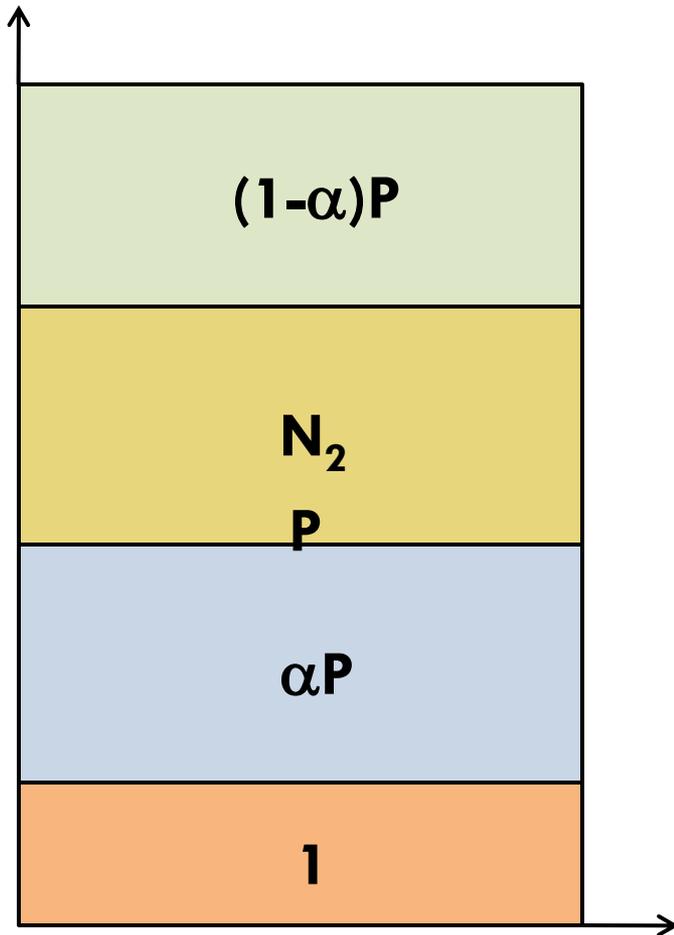
Gaussian Broadcast Channel



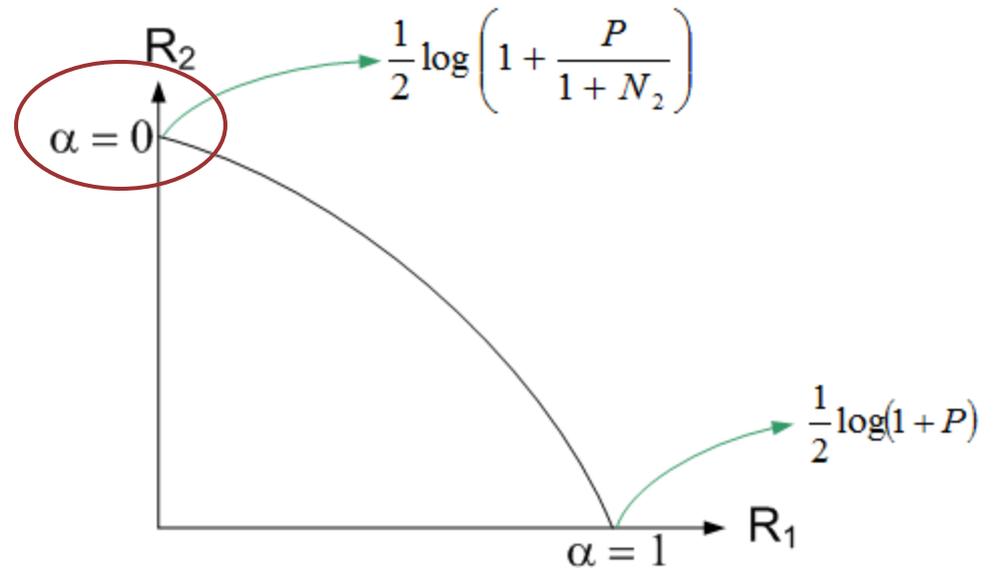
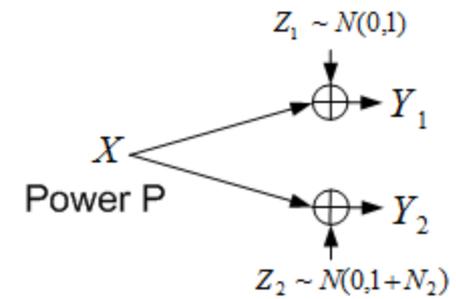
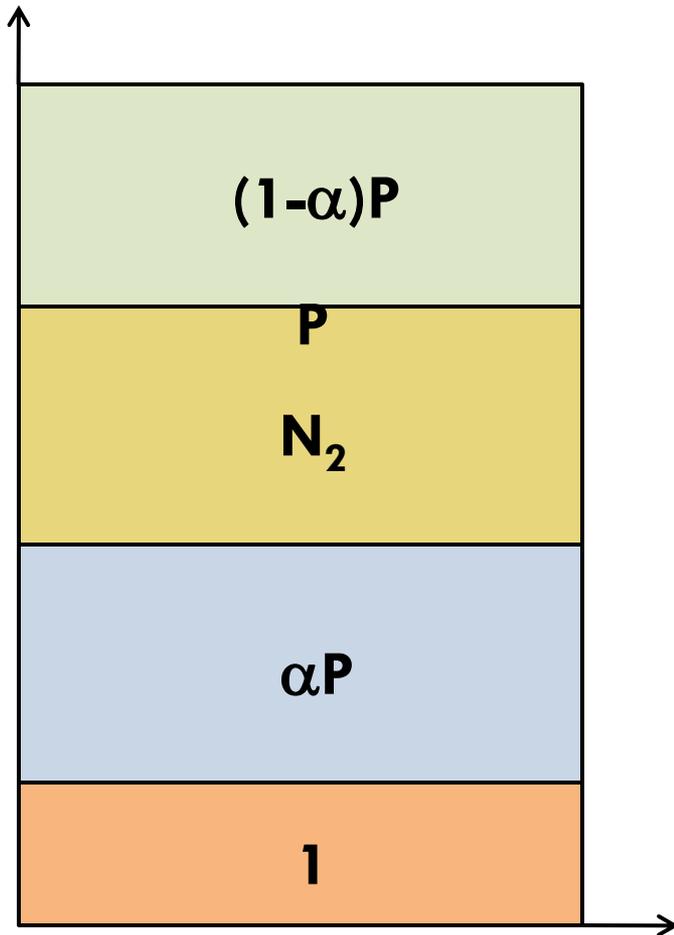
$$C_{BC} \{R_1, R_2\} : \quad 0 \leq R_1 \leq \frac{1}{2} \log(1 + \alpha P)$$

$$0 \leq \alpha \leq 1 \quad 0 \leq R_2 \leq \frac{1}{2} \log \left(1 + \frac{(1 - \alpha)P}{1 + N_2 + \alpha P} \right)$$

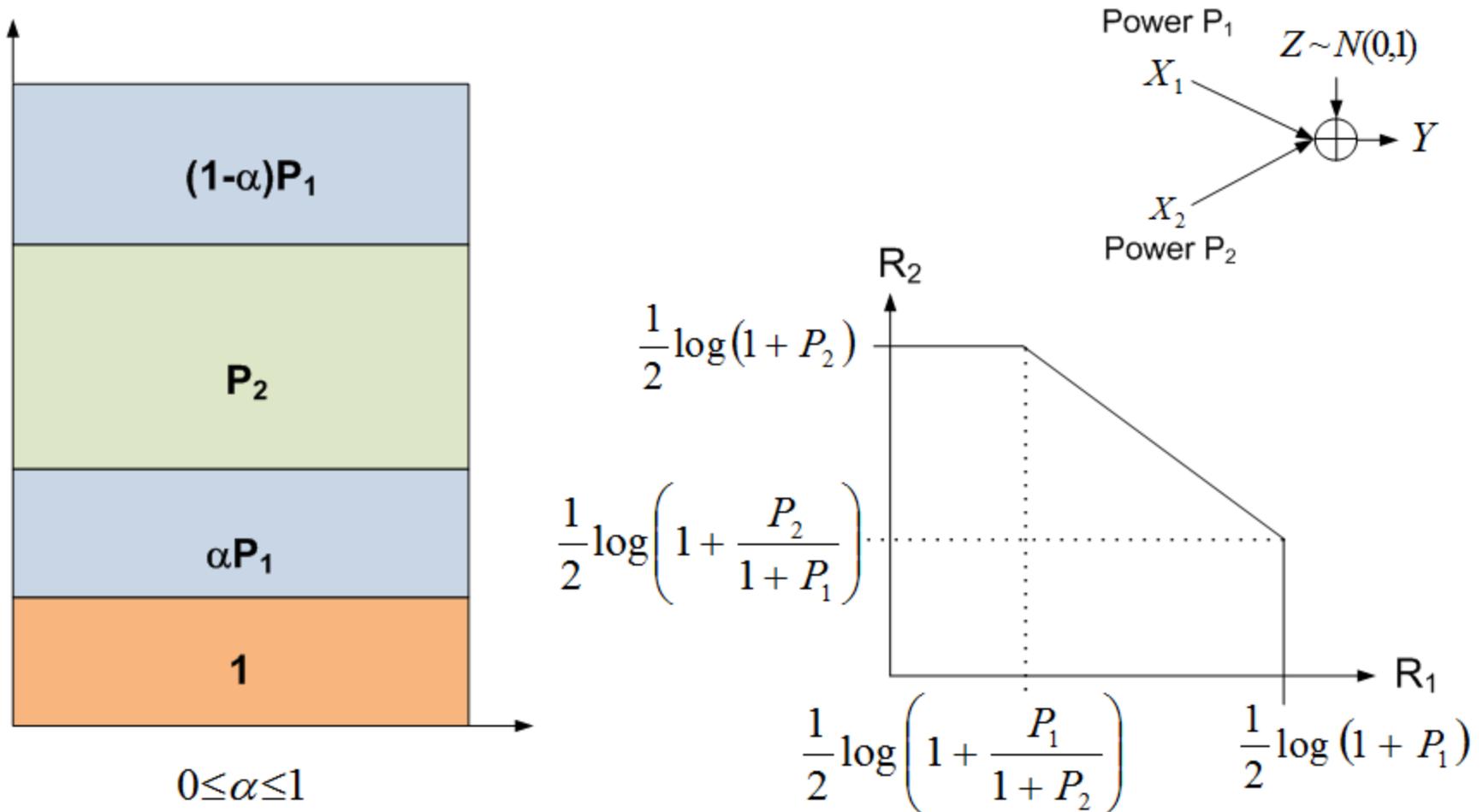
Superposition coding



Superposition coding

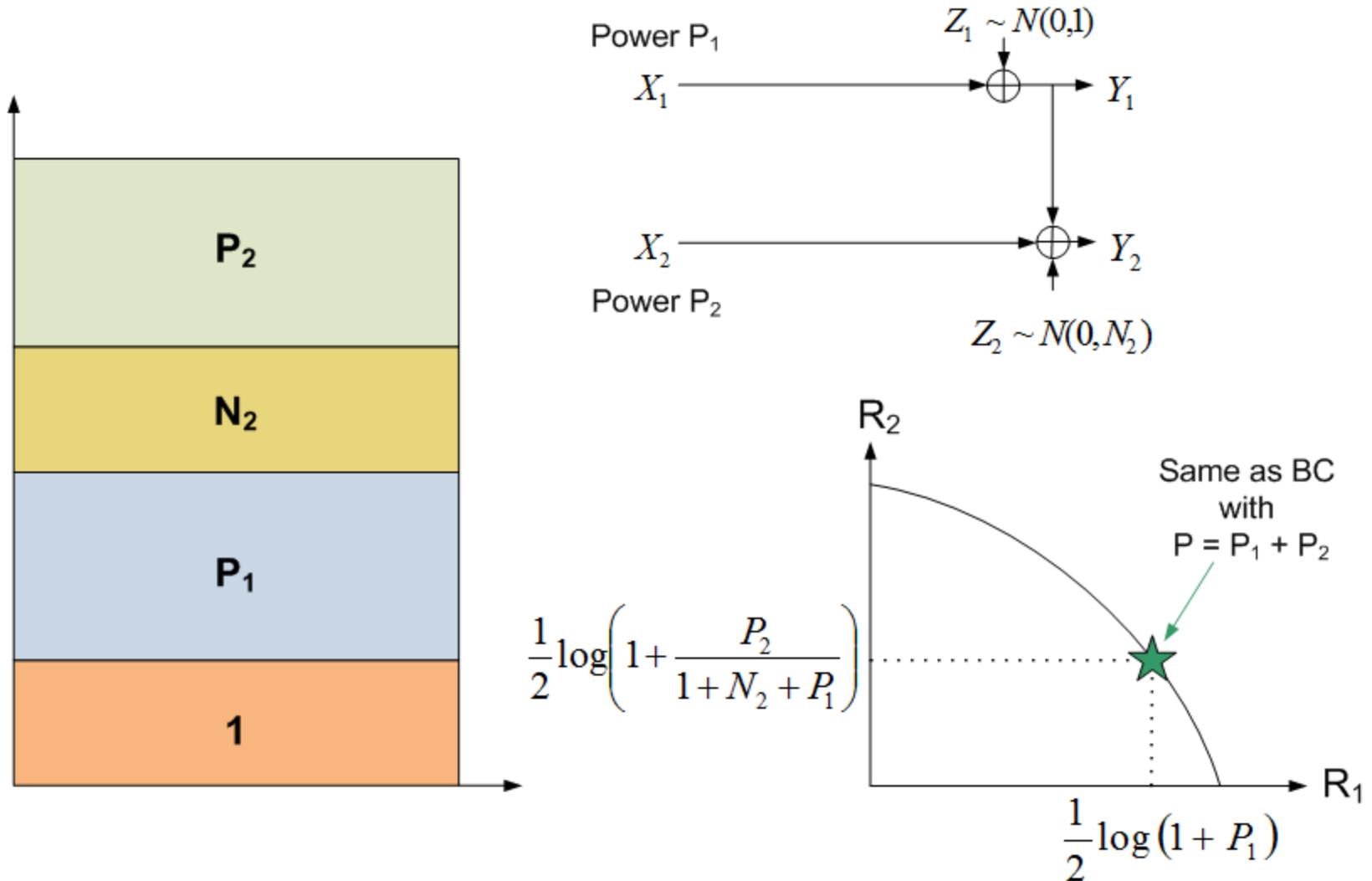


Multiple Access Channel



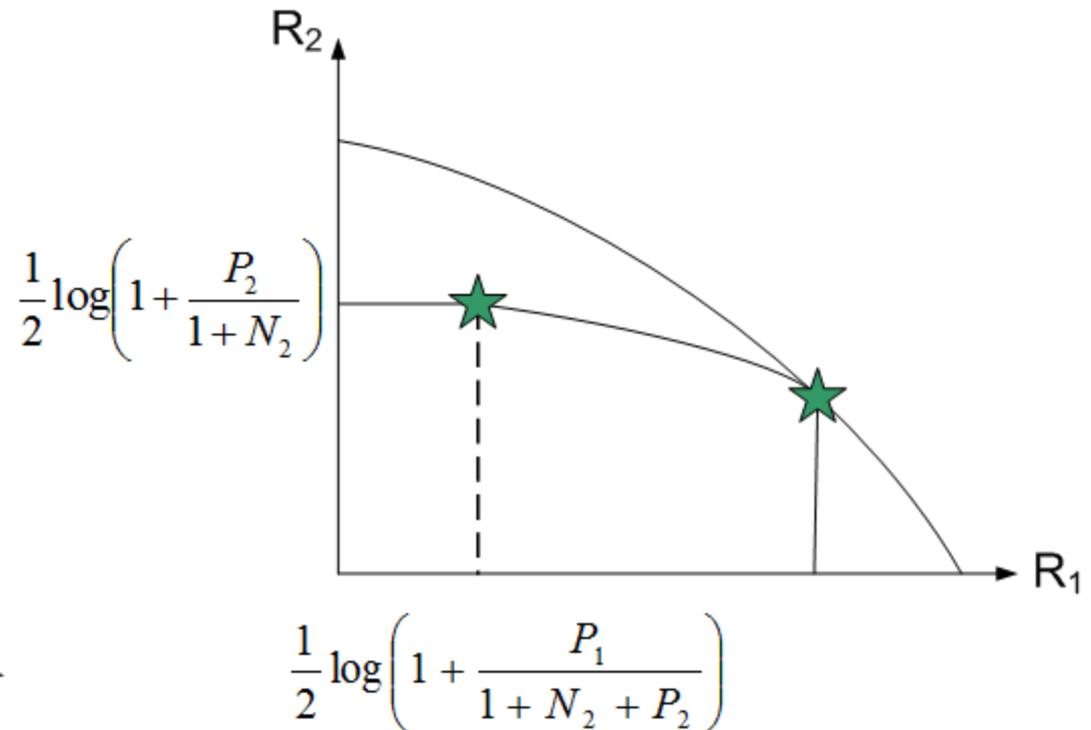
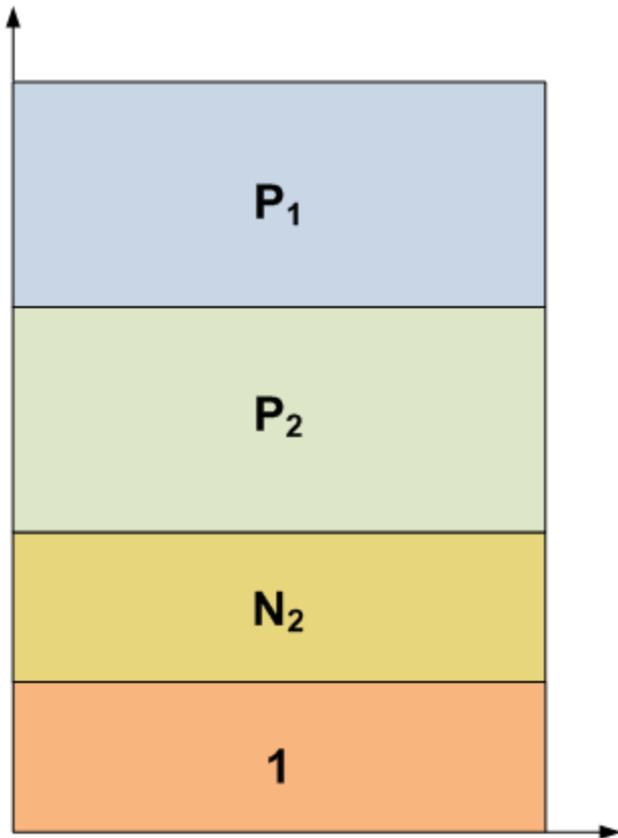
Degraded Interference Channel

- One Extreme Point

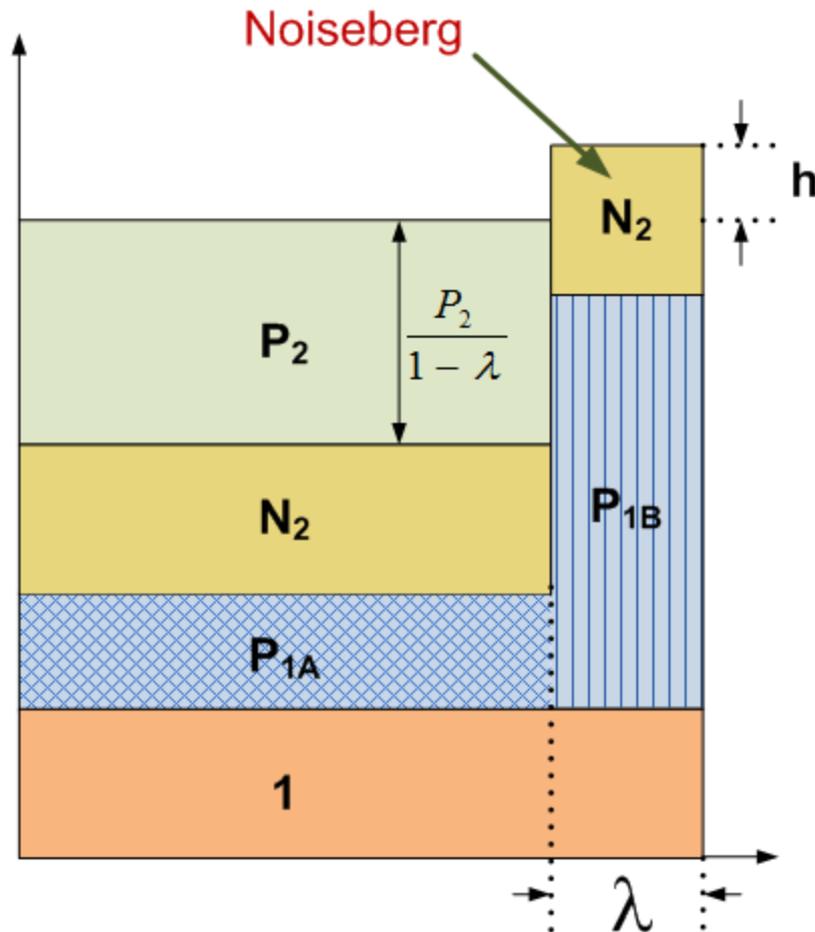


Degraded Interference Channel

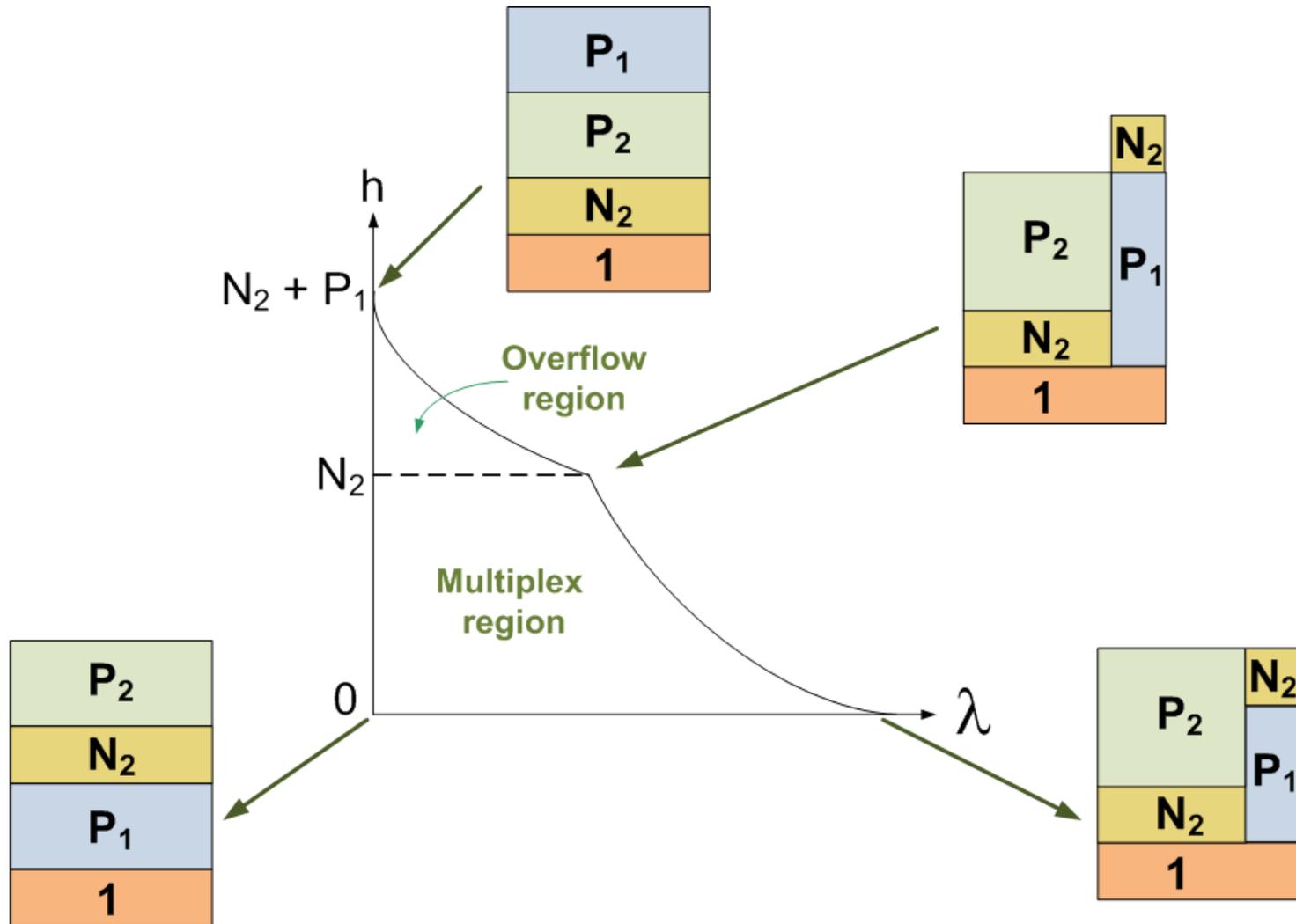
- Another Extreme Point



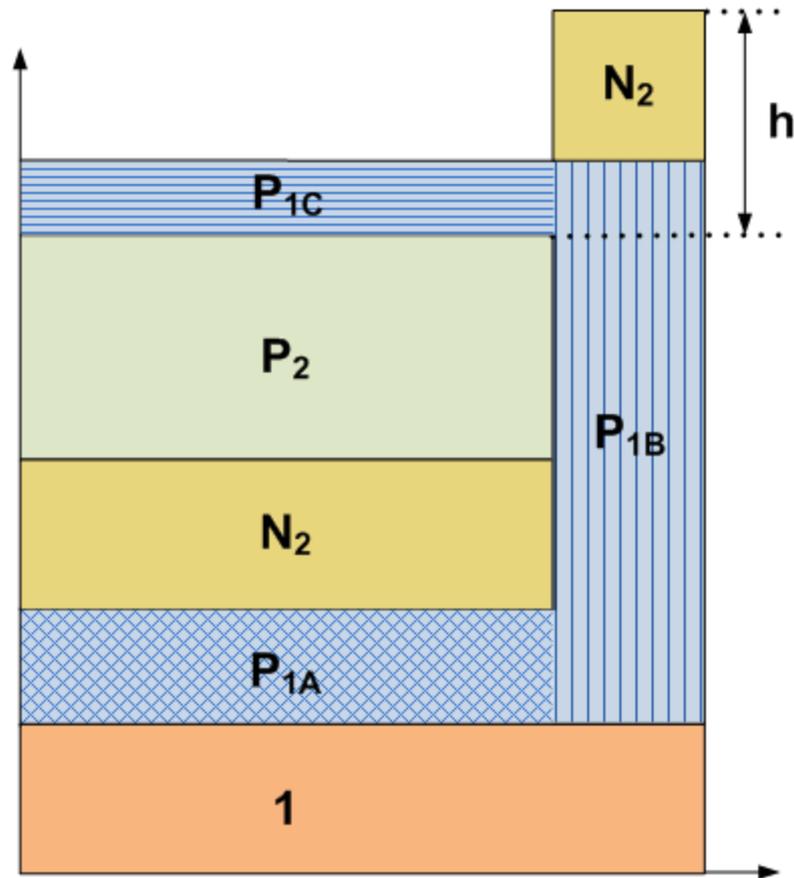
Intermediary Points (Multiplex Region)



Admissible region for (λ, h)

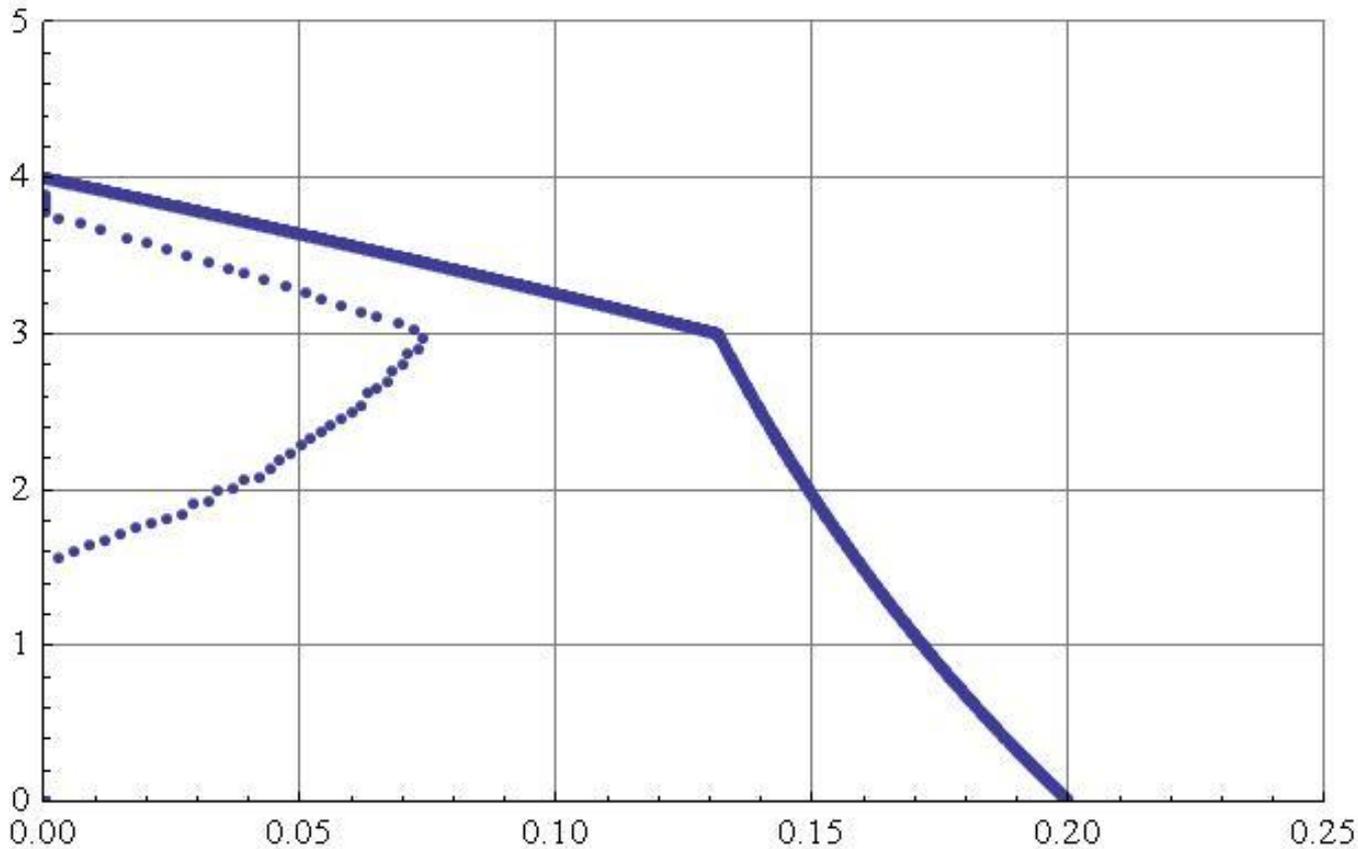


Intermediary Point (Overflow Region)



Admissible region

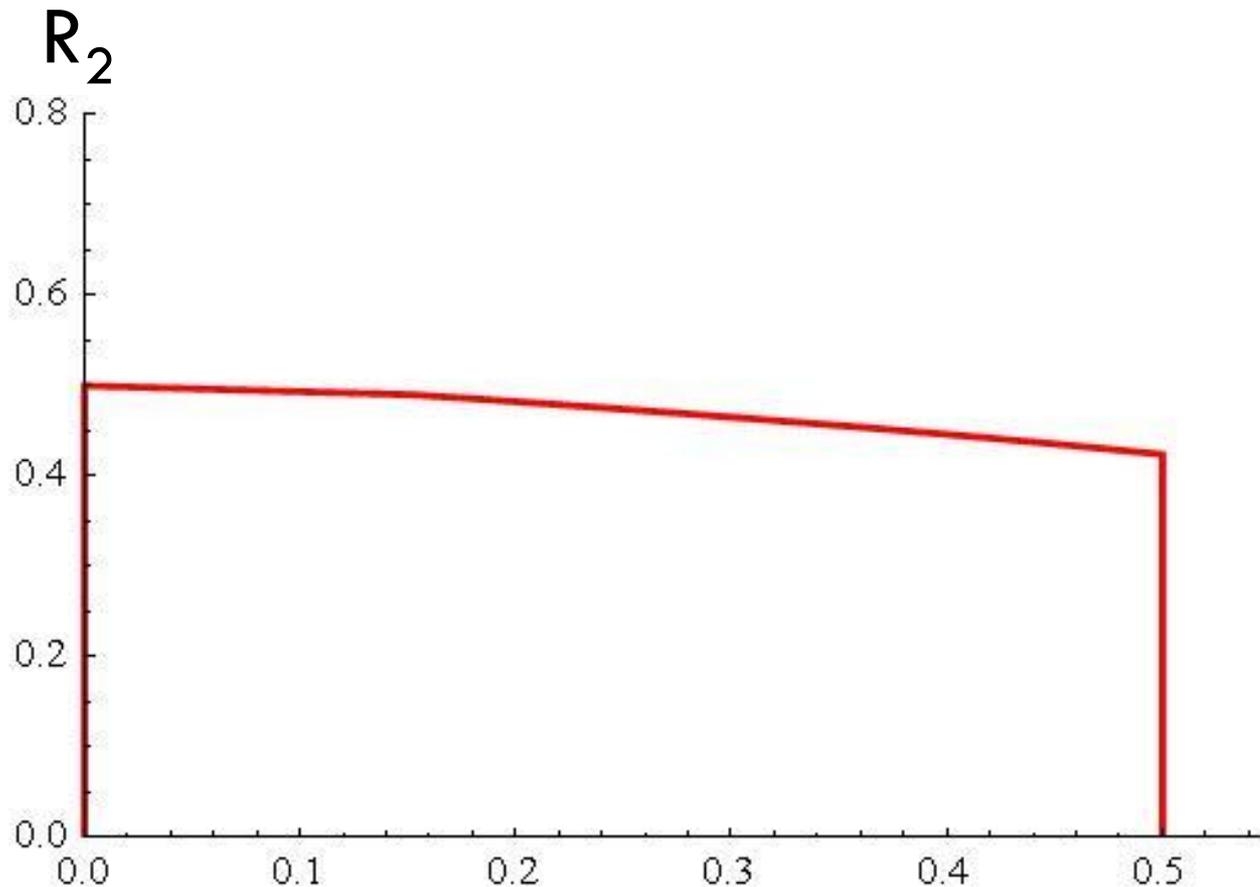
h



$$Q_1 = 1$$
$$Q_2 = 1$$
$$\alpha = 0.5$$
$$N_2 = 3$$

λ

The Z-Gaussian Interference Channel Rate Region

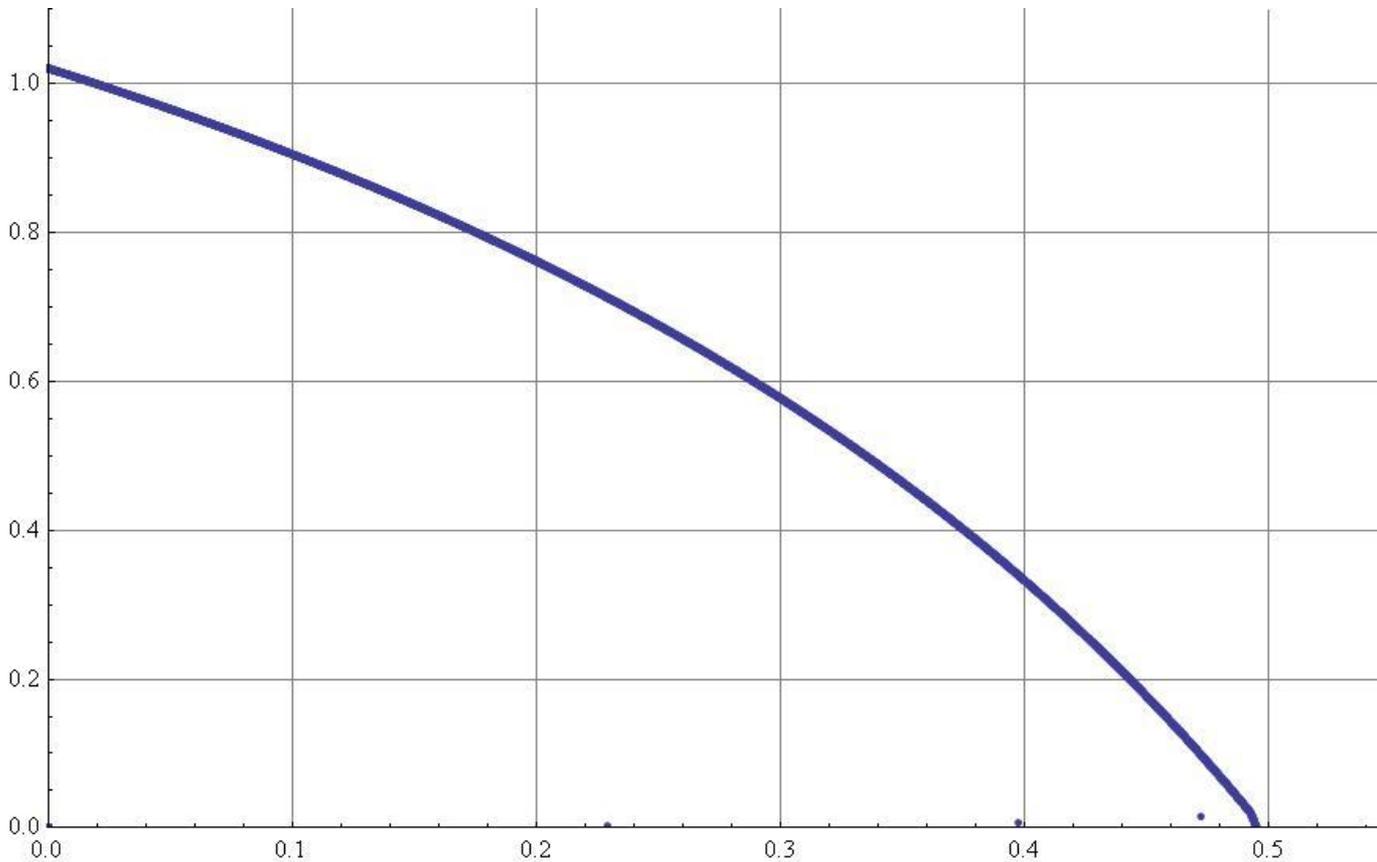


$$Q_1 = 1$$
$$Q_2 = 1$$
$$\alpha = 0.5$$
$$N_2 = 3$$

R_1

Admissible region

h



$$Q_1 = 1$$

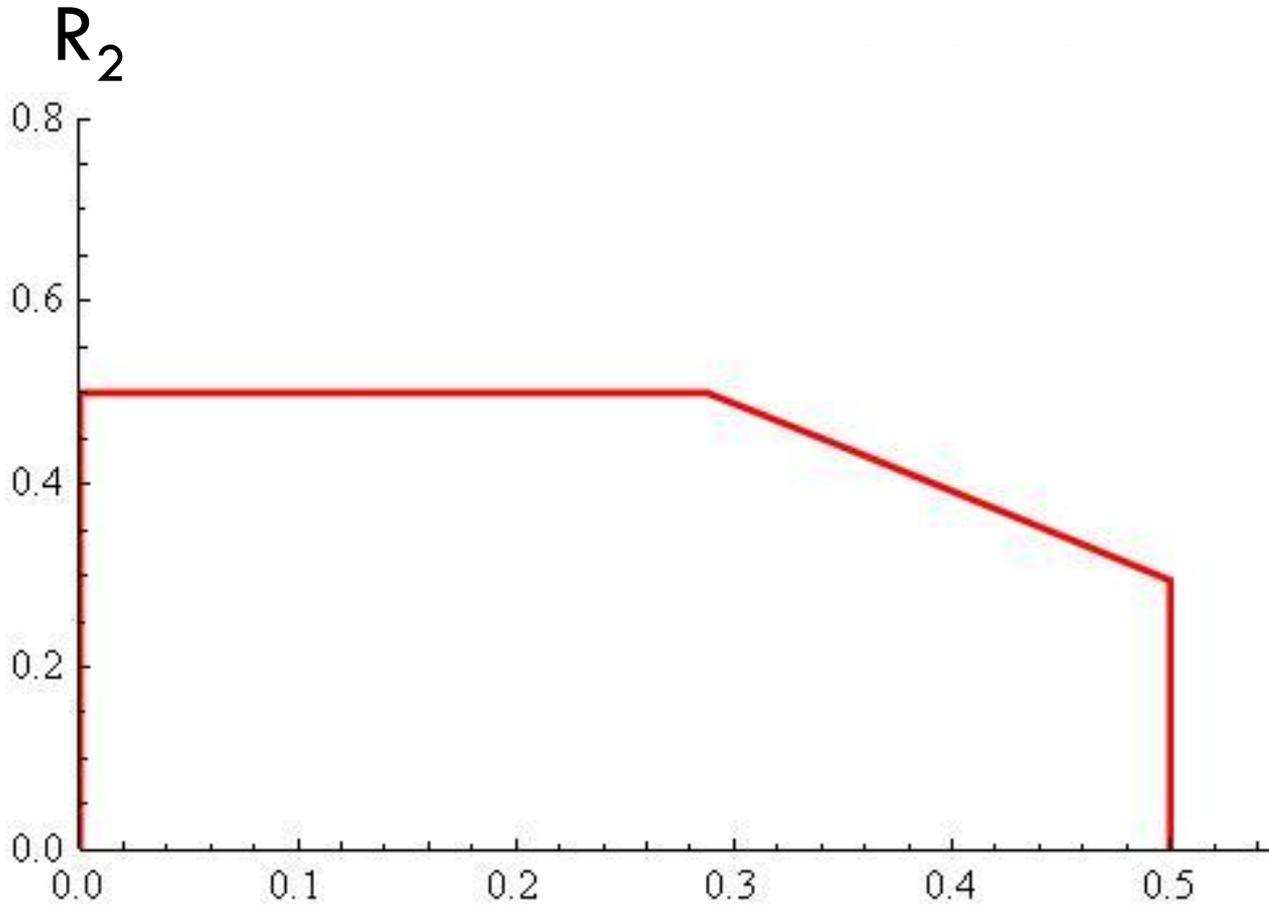
$$Q_2 = 1$$

$$\alpha = 0.99$$

$$N_2 = 0.02$$

λ

The Z-Gaussian Interference Channel Rate Region



$$Q_1 = 1$$

$$Q_2 = 1$$

$$\alpha = 0.99$$

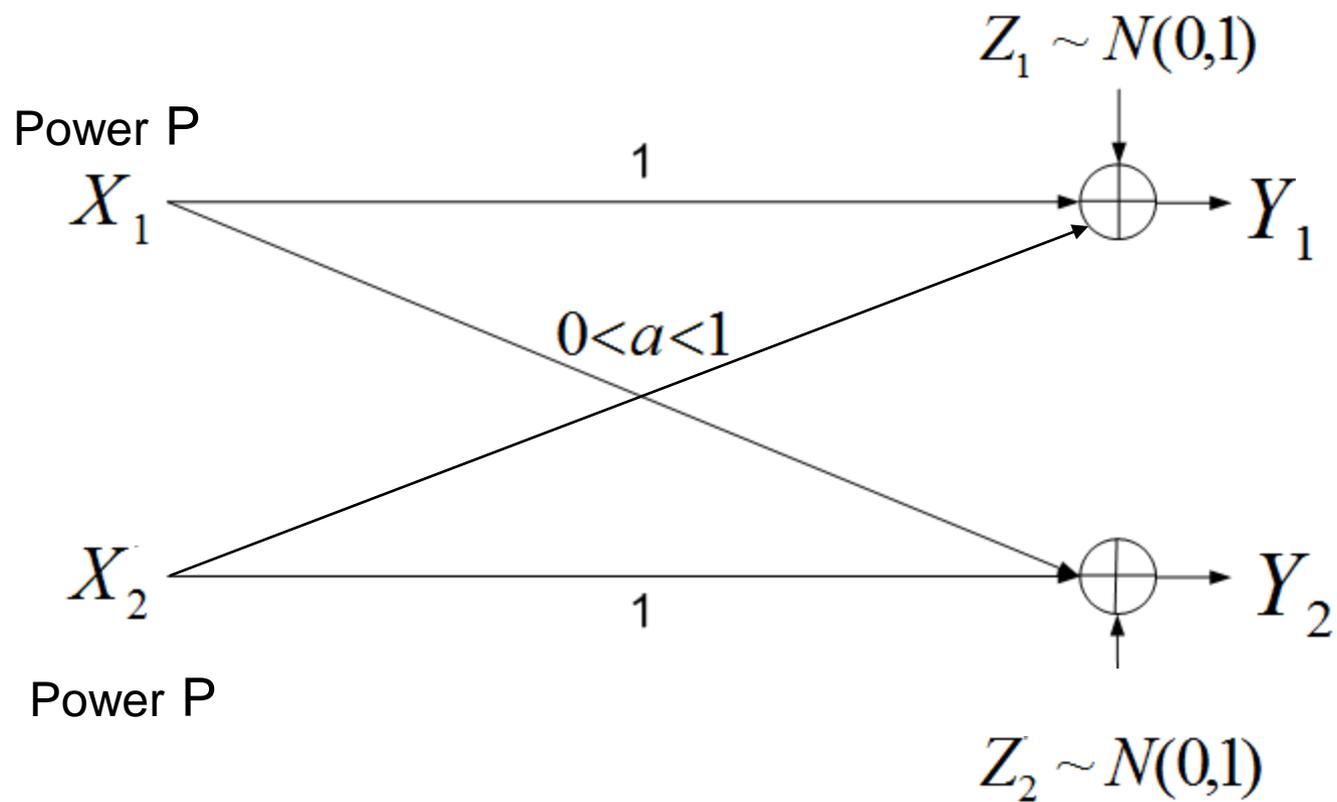
$$N_2 = 0.02$$

R_1

Remarks

- This is Han-Kobayashi region for Gaussian signaling (Zhao et al., ISIT-2012)
- Simple 2-D parameter space: (λ, h)
- Need entropy power-like inequality to establish capacity region

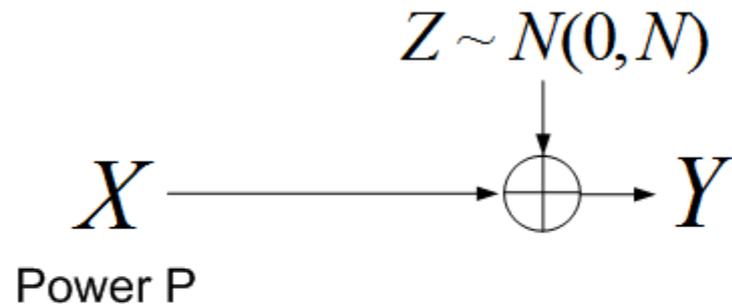
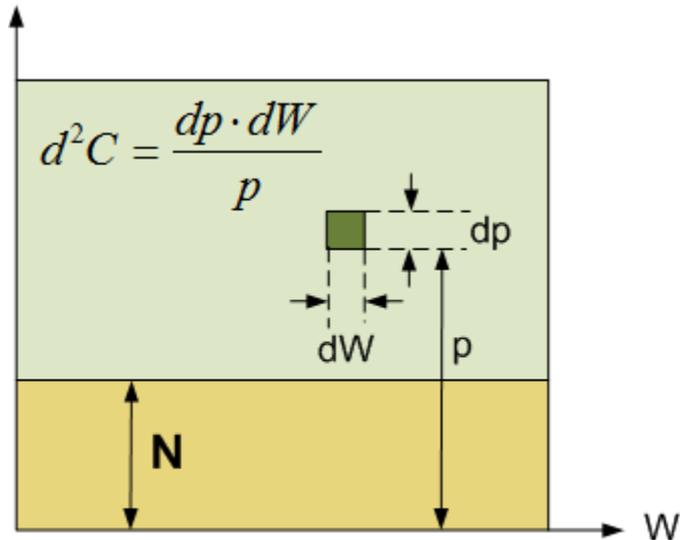
Symmetric Gaussian Interference Channel



Symmetric Interference Channels

- Discrete time channel seen as a band limited channel – differential capacity
- Concave envelopes
- Symmetric and Asymmetric Superposition
- Phase transitions in parameter space

Differential capacity

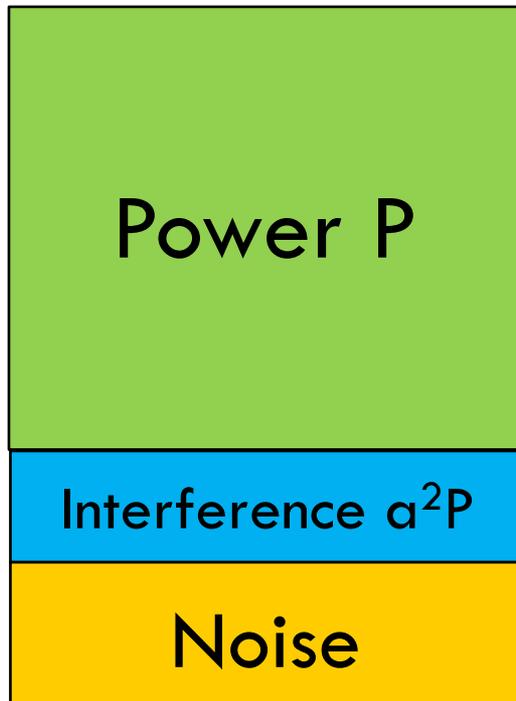


$$C = \iint d^2C = \frac{1}{2} \log \left(1 + \frac{P}{N} \right)$$

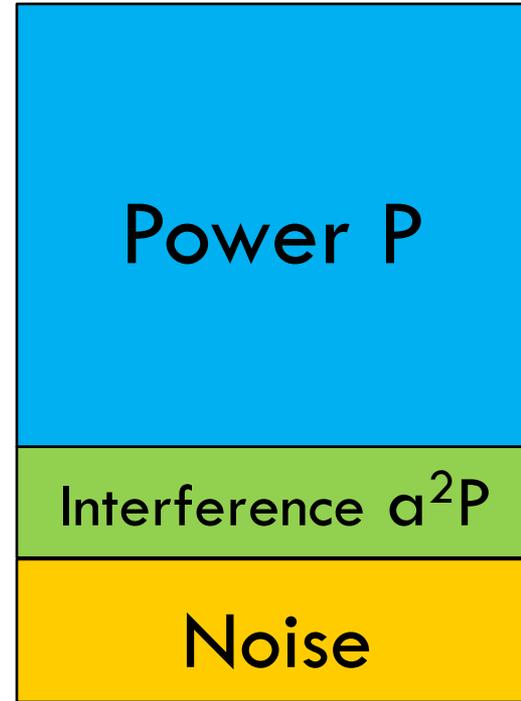
Discrete time channel seen as a band limited channel

Interference channel: Spectra at Y_1 and Y_2

□ At Y_1

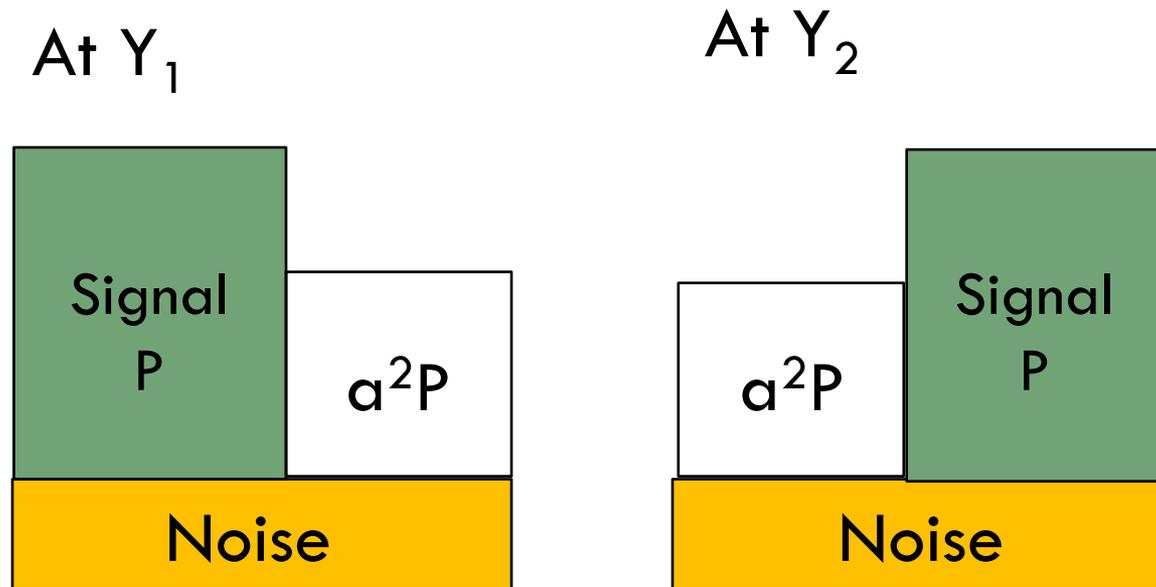


At Y_2



IAN: $R_1 + R_2 \leq \log \left(1 + \frac{P}{1 + a^2P} \right)$

Interference Channel: TDM/FDM:

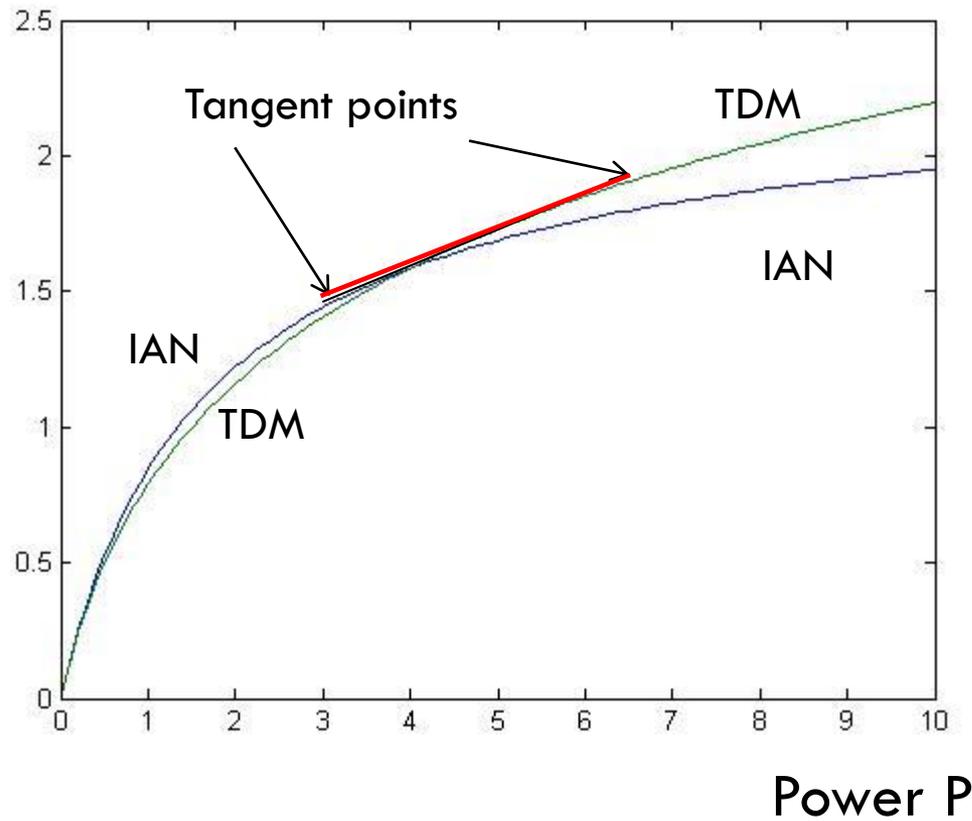


$$R_1 + R_2 \leq \frac{1}{2} \log(1 + 2P)$$

Concave Envelope

IAN vs TDM/FDM, $\alpha^2=0.25$

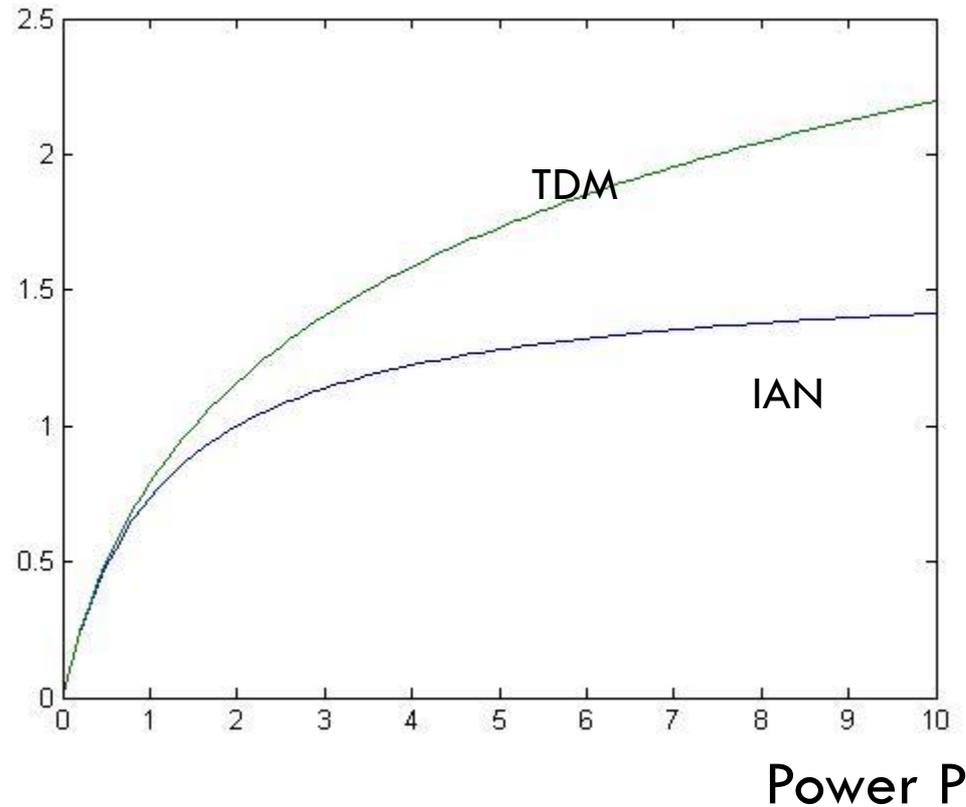
Rate Sum



Multiplex domination

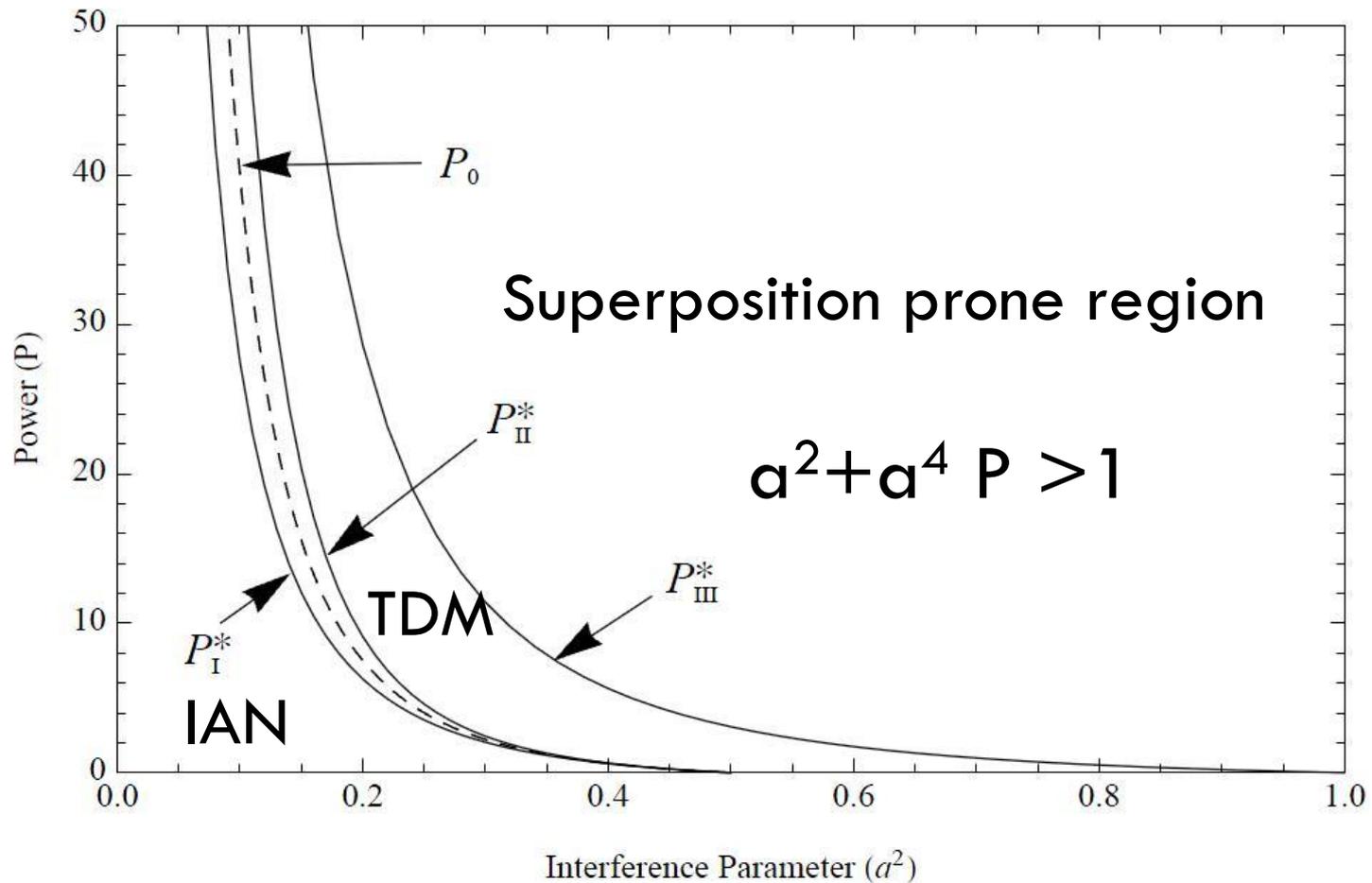
IAN vs TDM/FDM, $\alpha^2=0.5$

Rate Sum

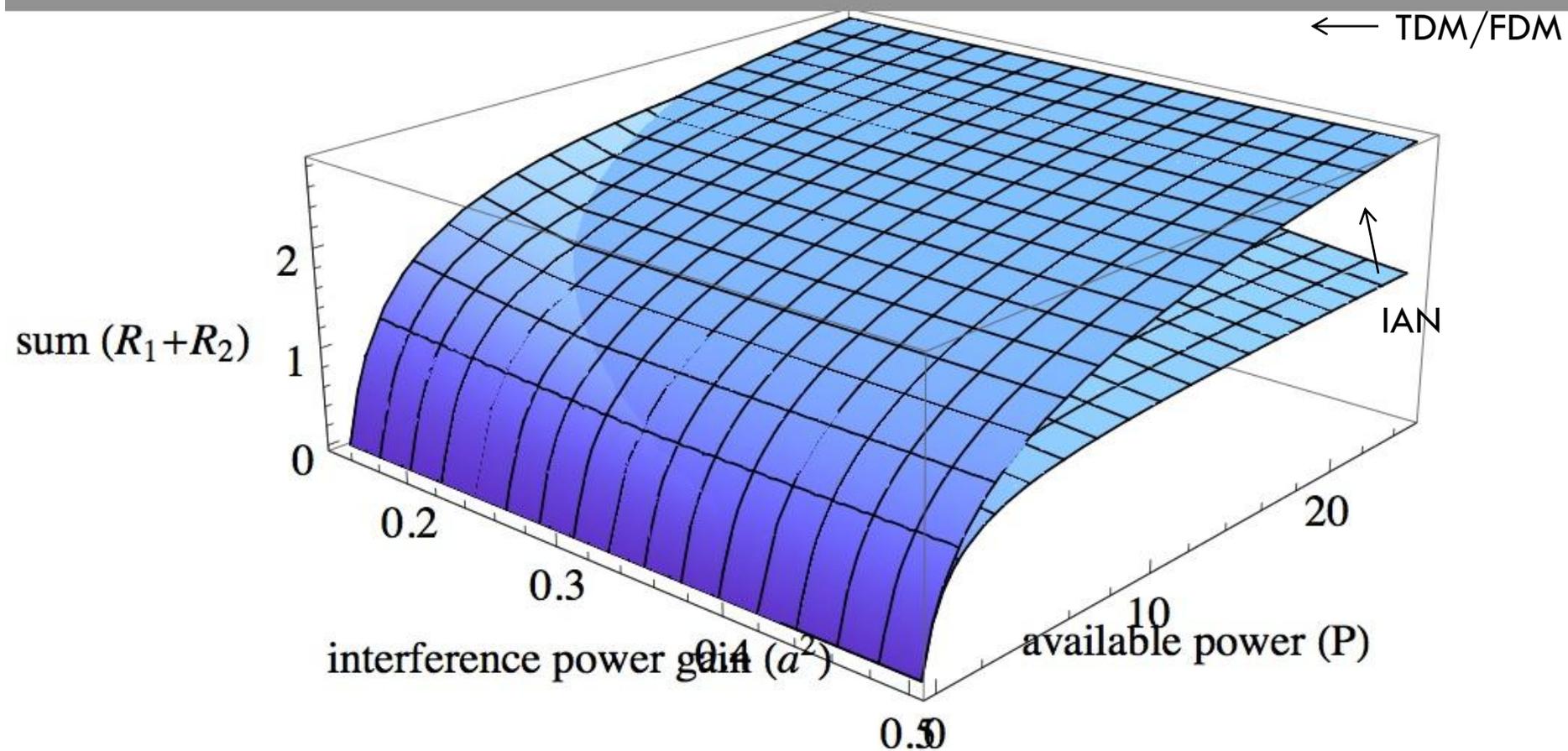


No intersection beyond $\alpha^2=0.5$

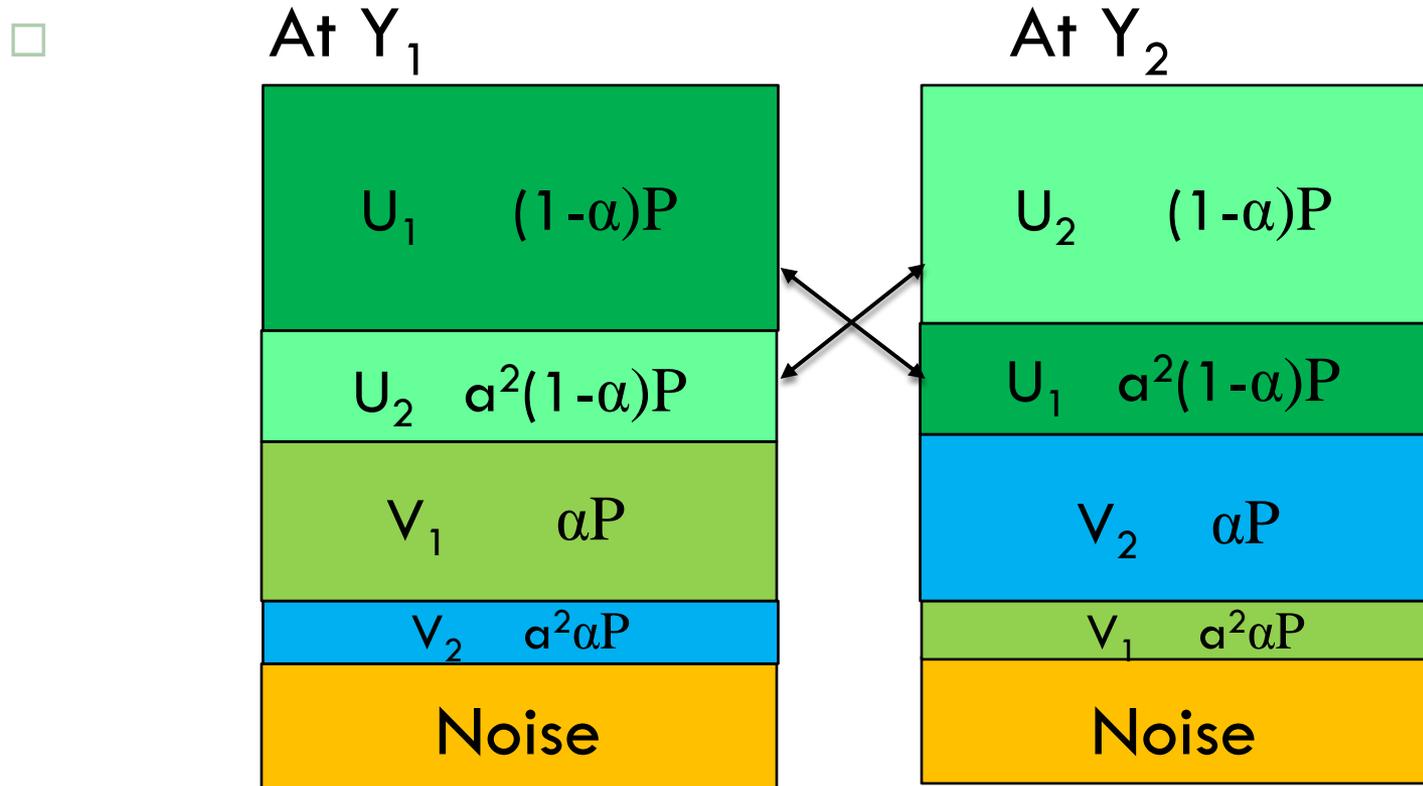
Interference as Noise and TDM/FDM



Rate Sum for IAN and TDM/FDM



Superposition: partially decoding

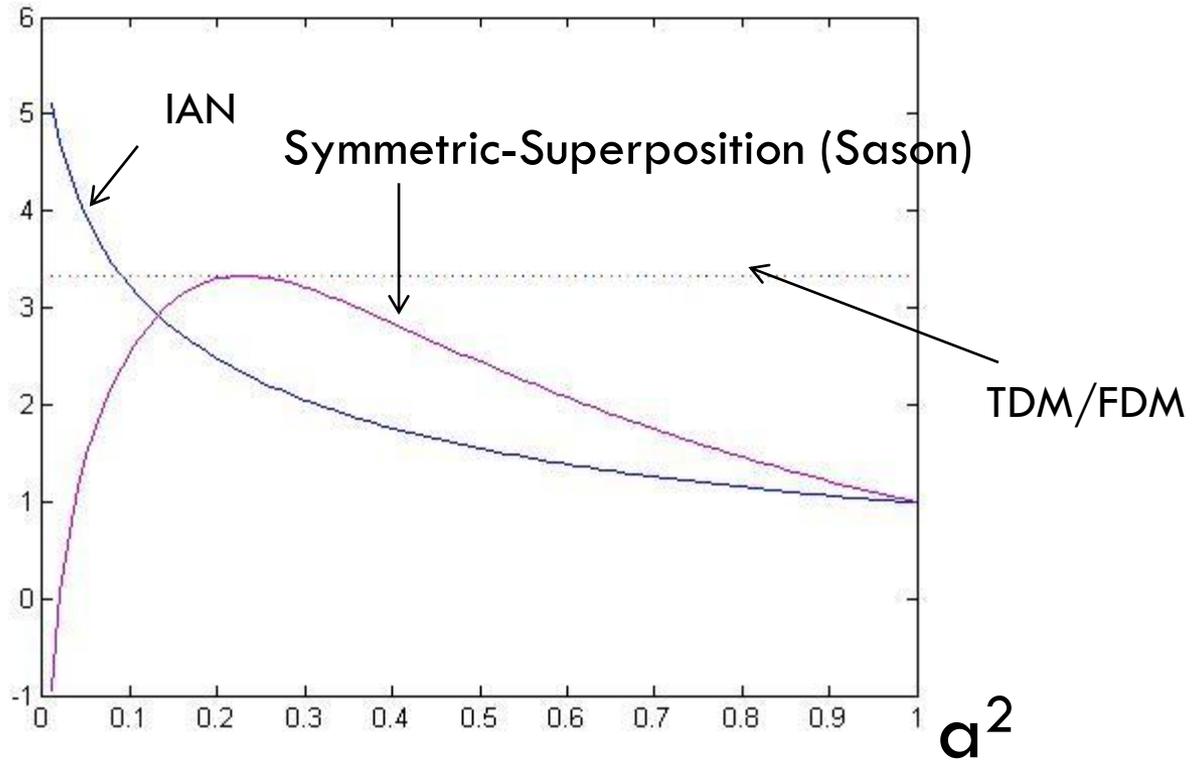


$$R_1 + R_2 \leq \log\left(\frac{1+P+a^2P}{\frac{1-a^2}{a^2} + a^2(1+a^2P)}\right), \text{ Sason (2004)}$$

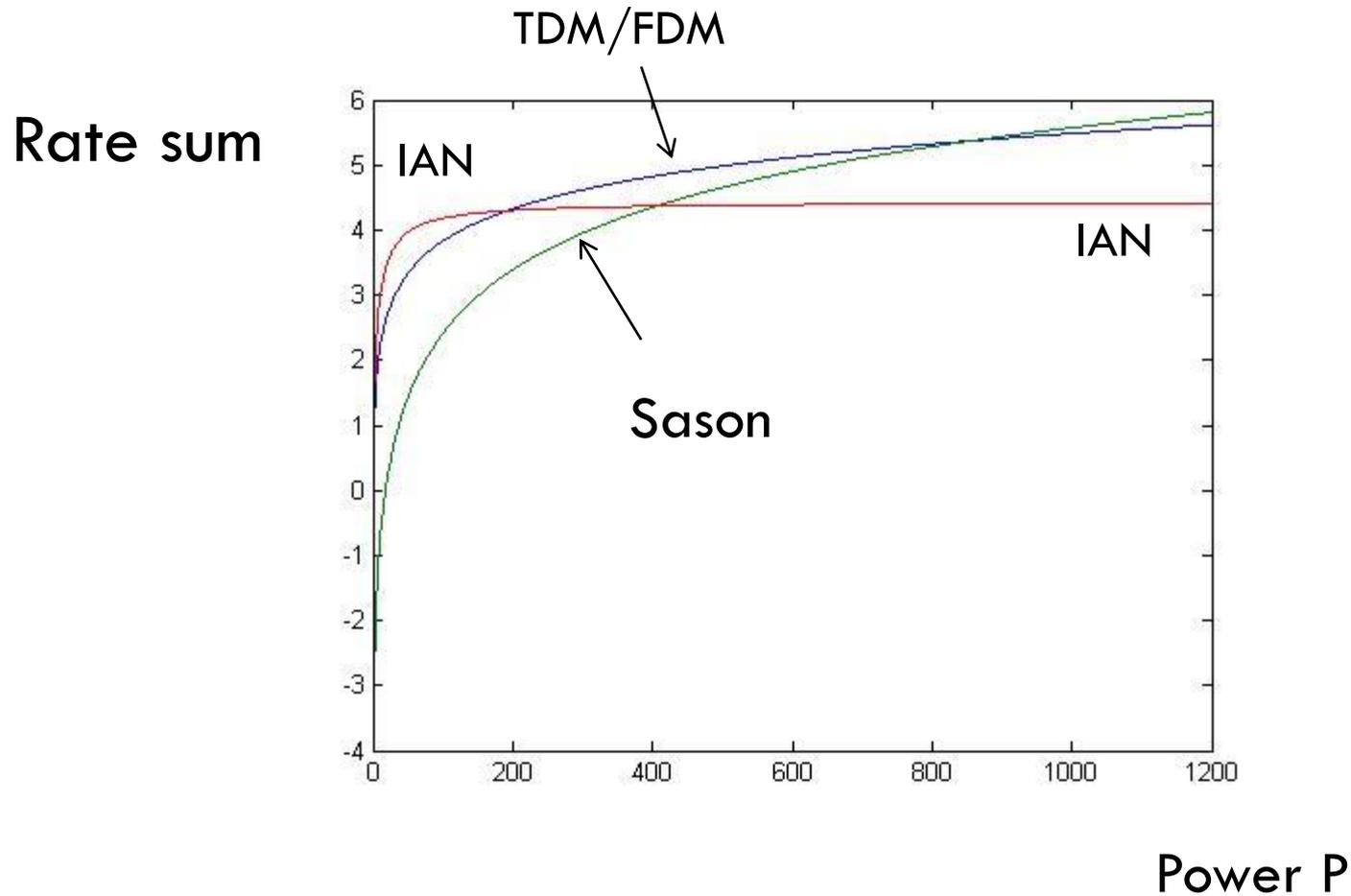
Point where Symmetric Superposition starts beating TDM/FDM

$P=50$

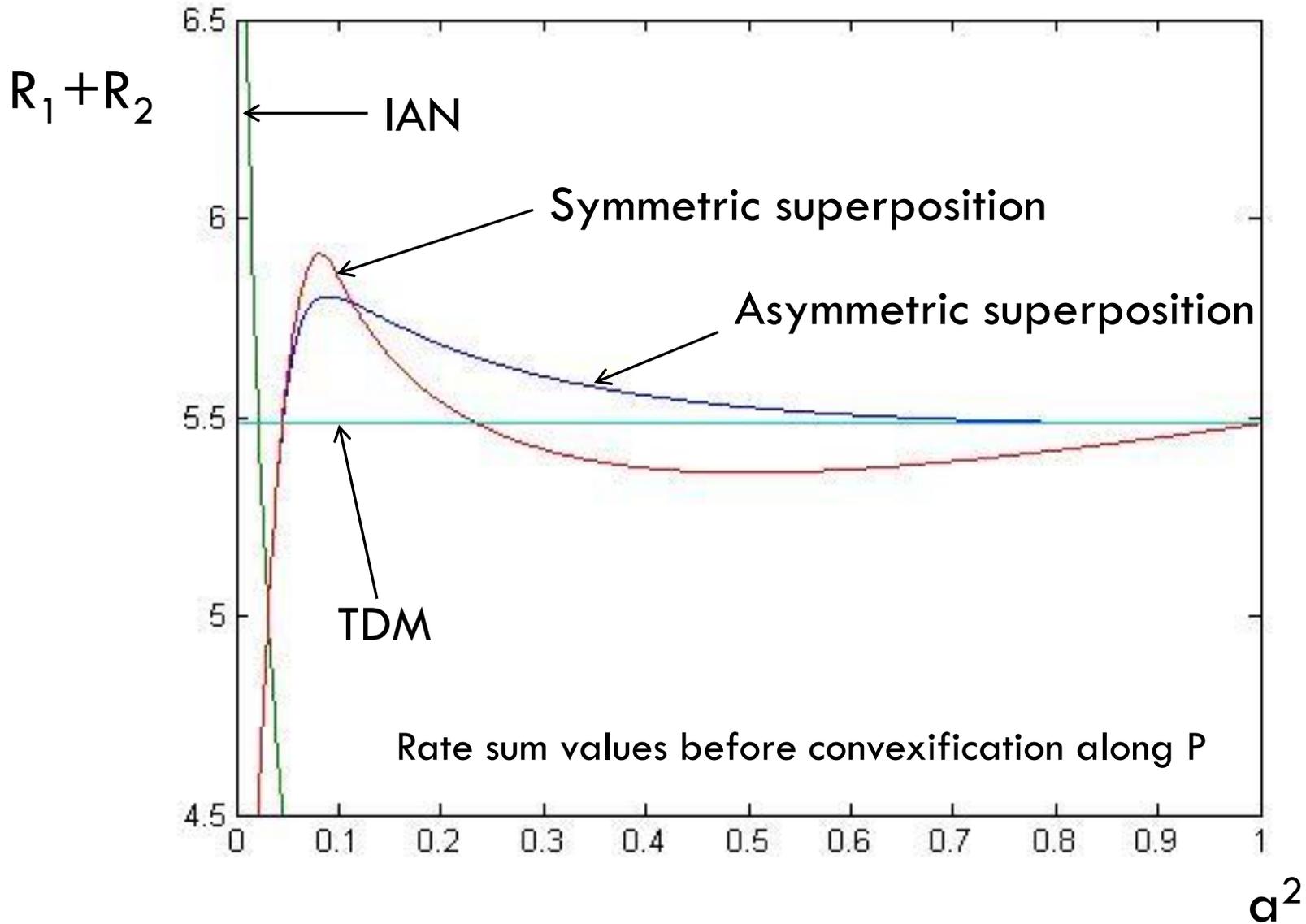
Rate Sum



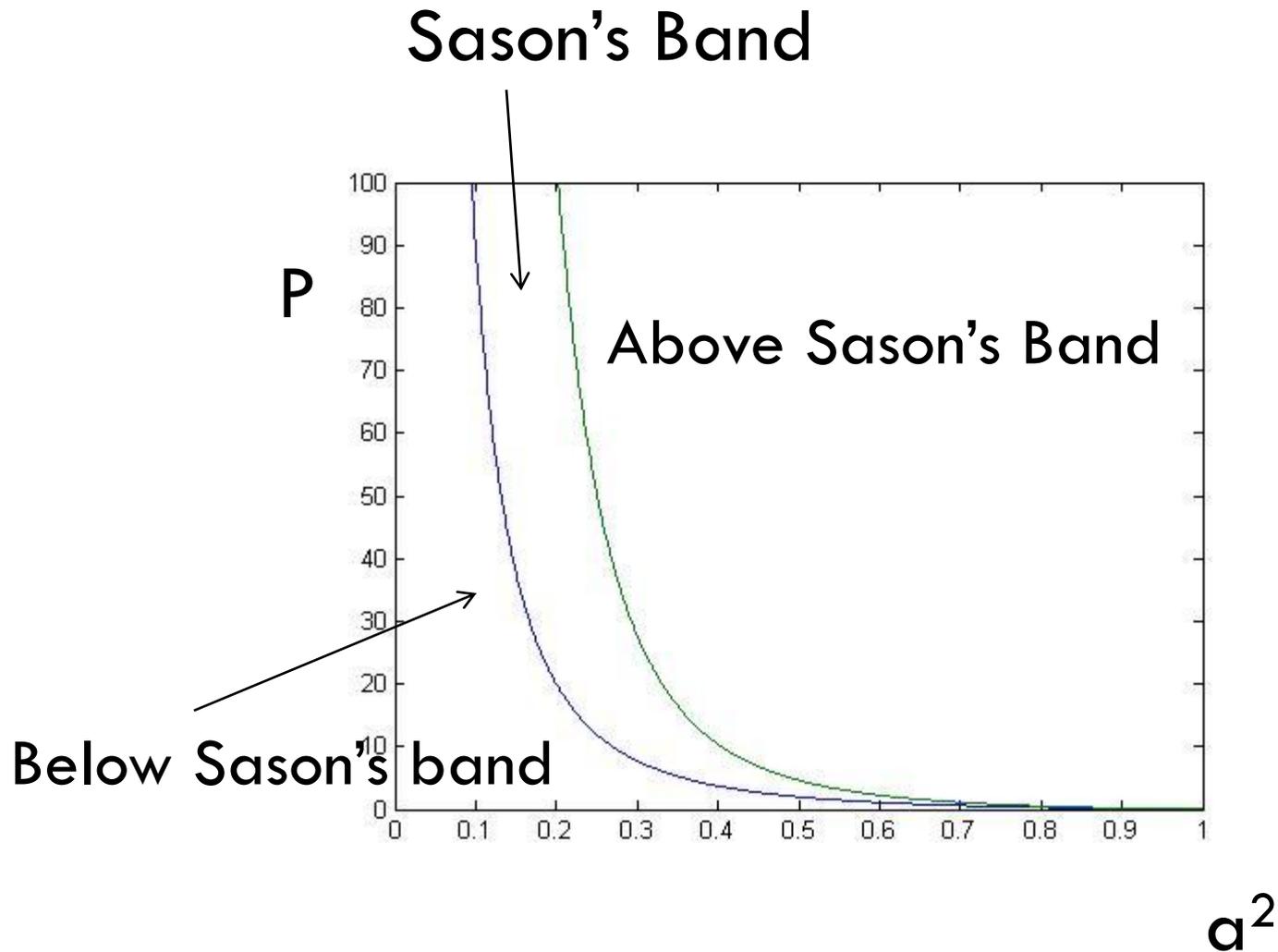
Rate Sum, $\alpha^2=0.05$: Need convexification



Rate sum for $P=1000$, $0 \leq a^2 \leq 1$



Symmetric superposition:



Symmetric Superposition (continued):

□ Optimal choice for $\alpha = \alpha_1 = \alpha_2$:

□ Case 1:

□ If $\frac{(1-a^2)}{a^4} \leq P \leq \frac{(1-a^6)}{a^6(1-a^2)}$ (*Sason's Band*)

then set $\alpha P = a^2(1 + a^2P) - 1$;

□ Case 2:

□ If $P \geq \frac{(1-a^6)}{a^6(1-a^2)}$ (*Above Sason's Band*)

then set $\alpha P = \frac{(1-a^2)}{a^2(1+a^2)}$.

Note: Invariant with P

Symmetric Superposition (continued):

□ In Sason's Band:

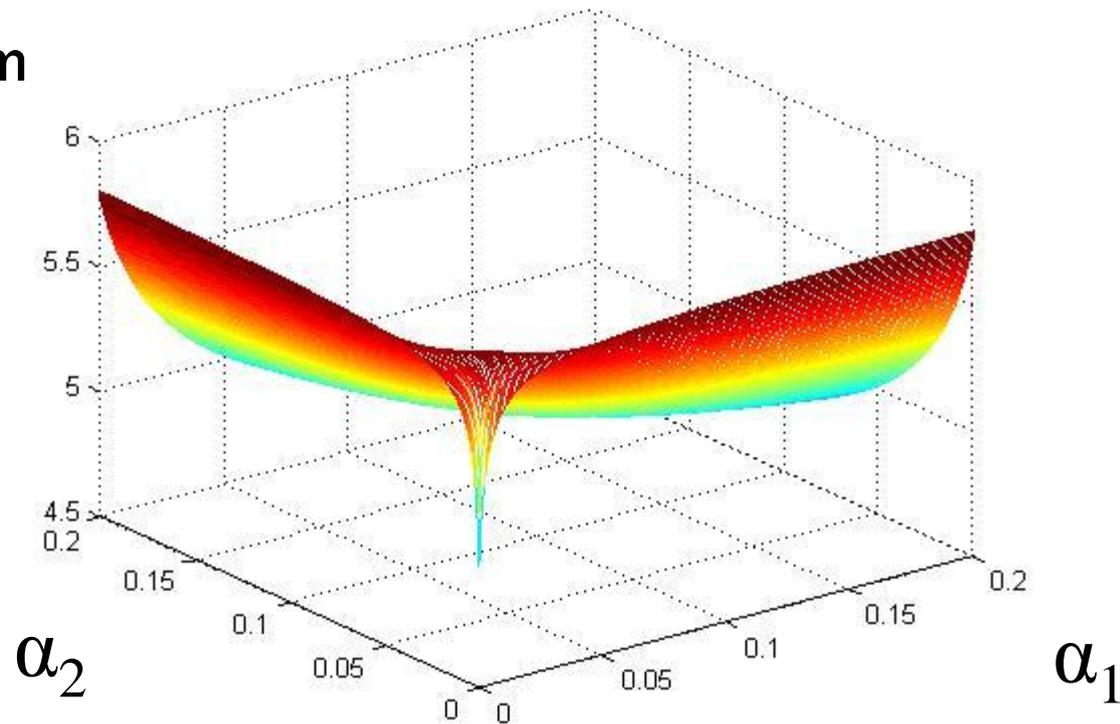
$$\square \quad R_1 + R_2 \leq \log \left(\frac{a^2(1+P+a^2P)}{1-a^2+a^4(1+a^2P)} \right)$$

□ Above Sason's Band:

$$\square \quad R_1 + R_2 \leq \frac{1}{2} \log \left(\frac{(1+a^2)^2(1+P+a^2P)}{4a^2} \right)$$

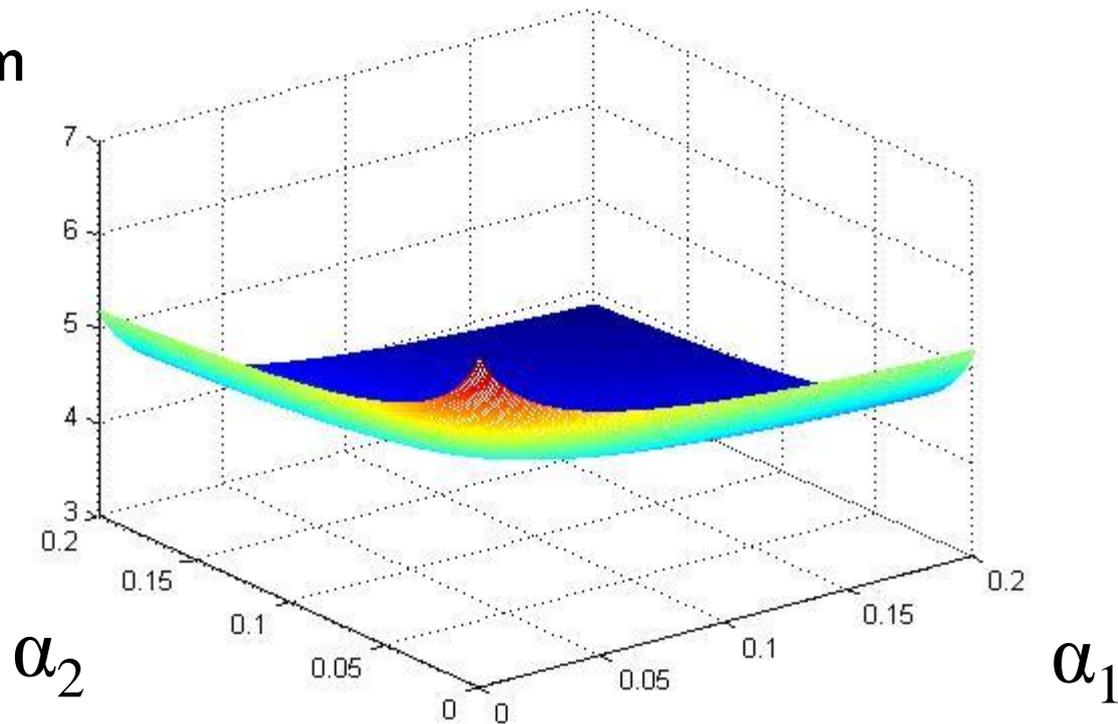
The hummingbird function:

Rate Sum



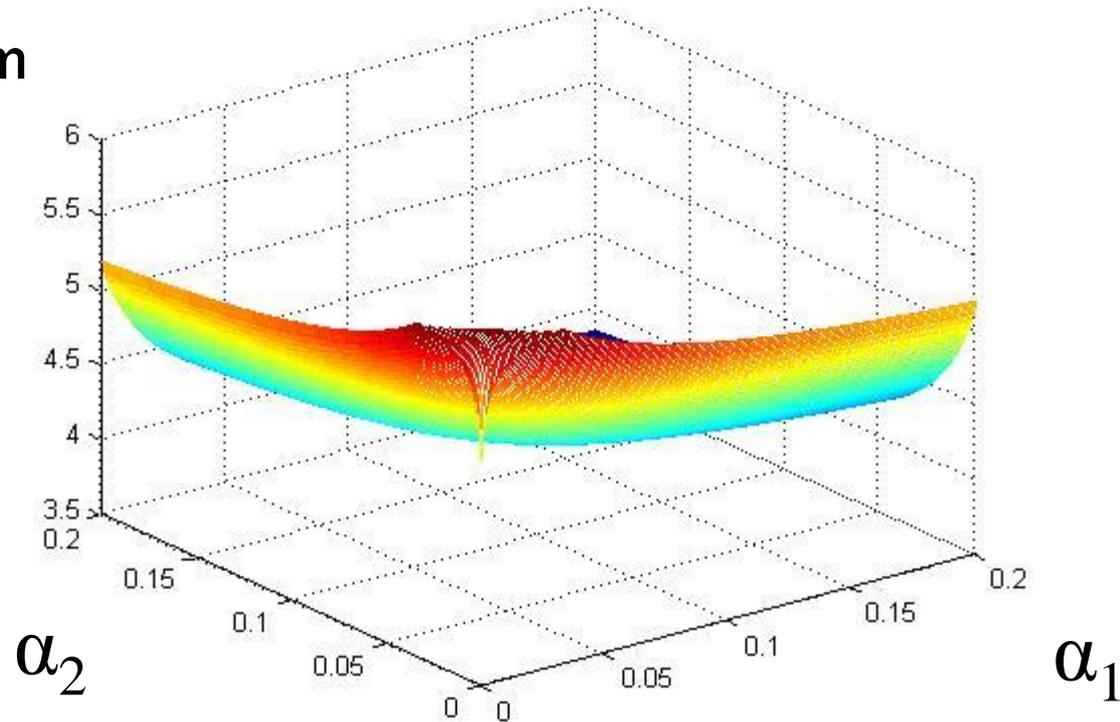
The shroud function

Rate Sum



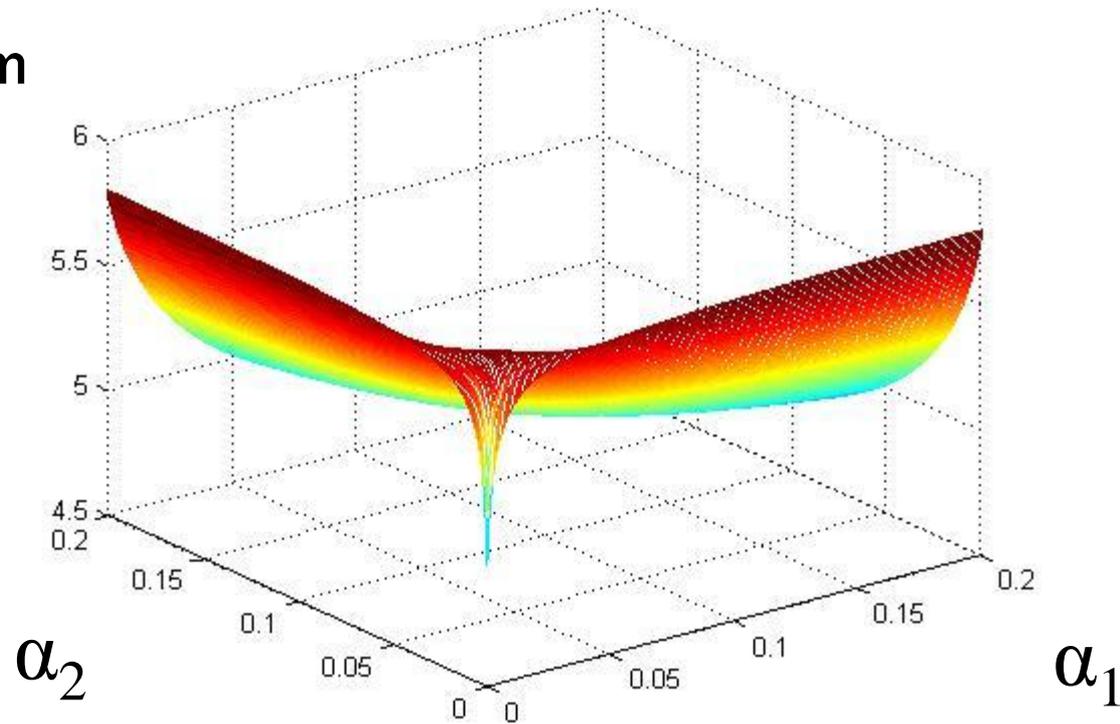
Min (hummingbird, shroud)

Rate Sum

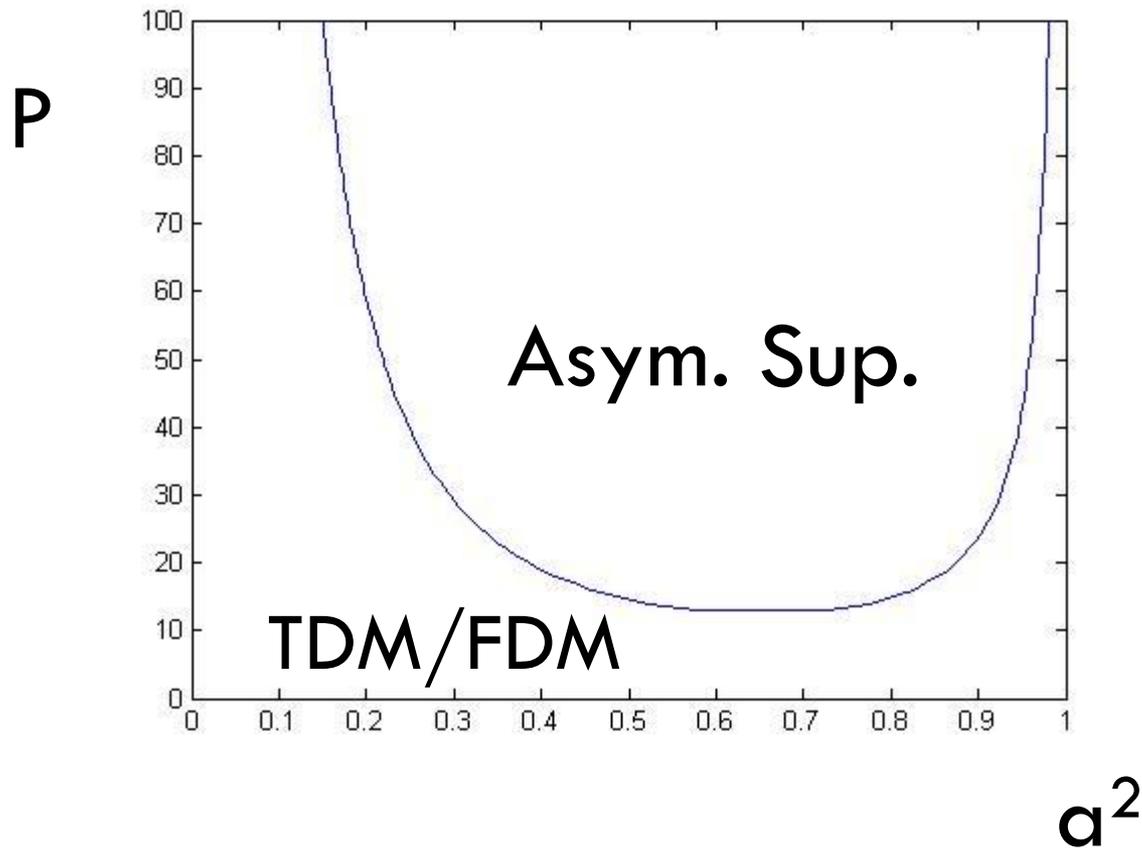


Flapping wings

Rate Sum

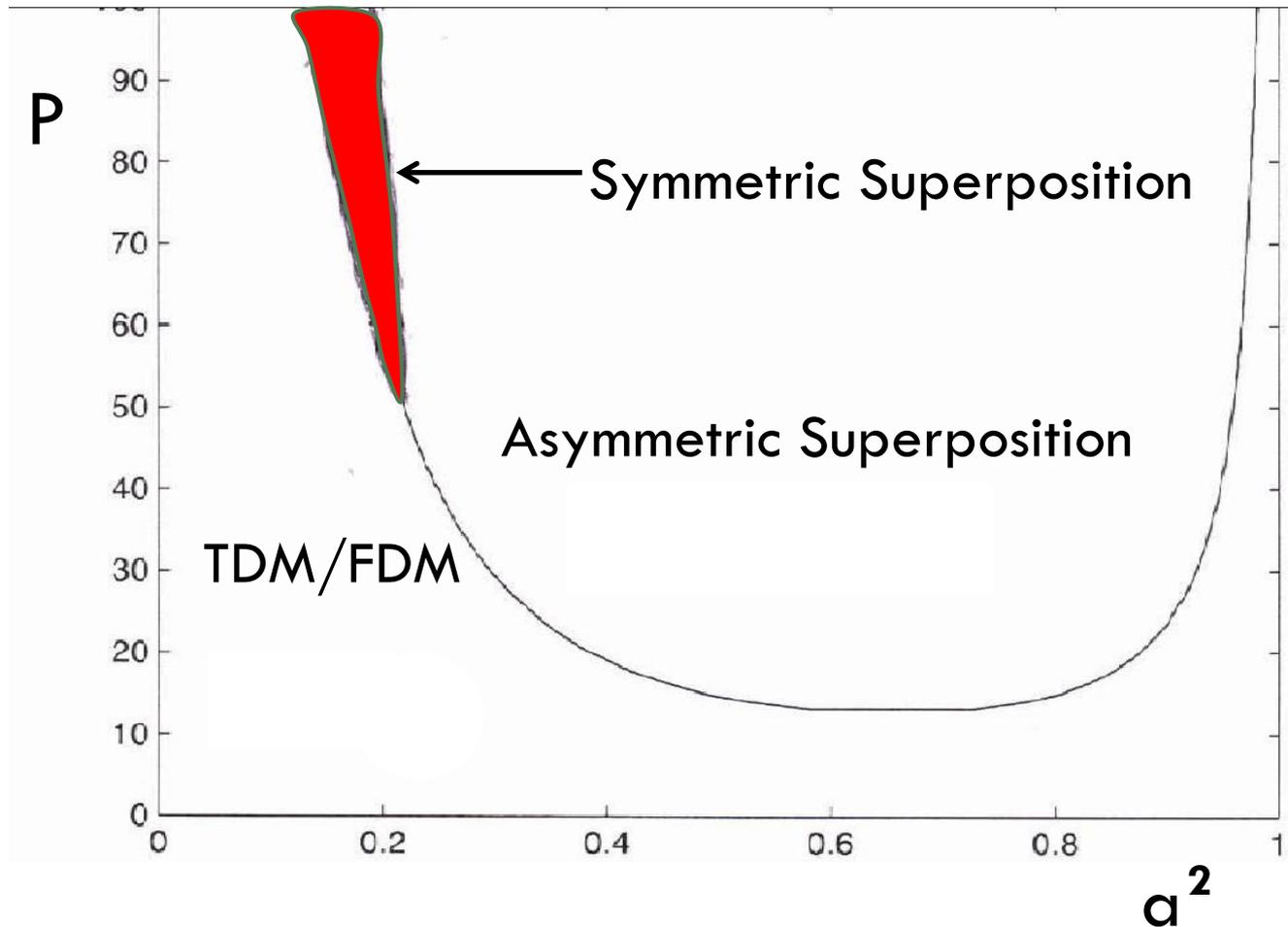


Asymmetric-Superposition vs TDM/FDM

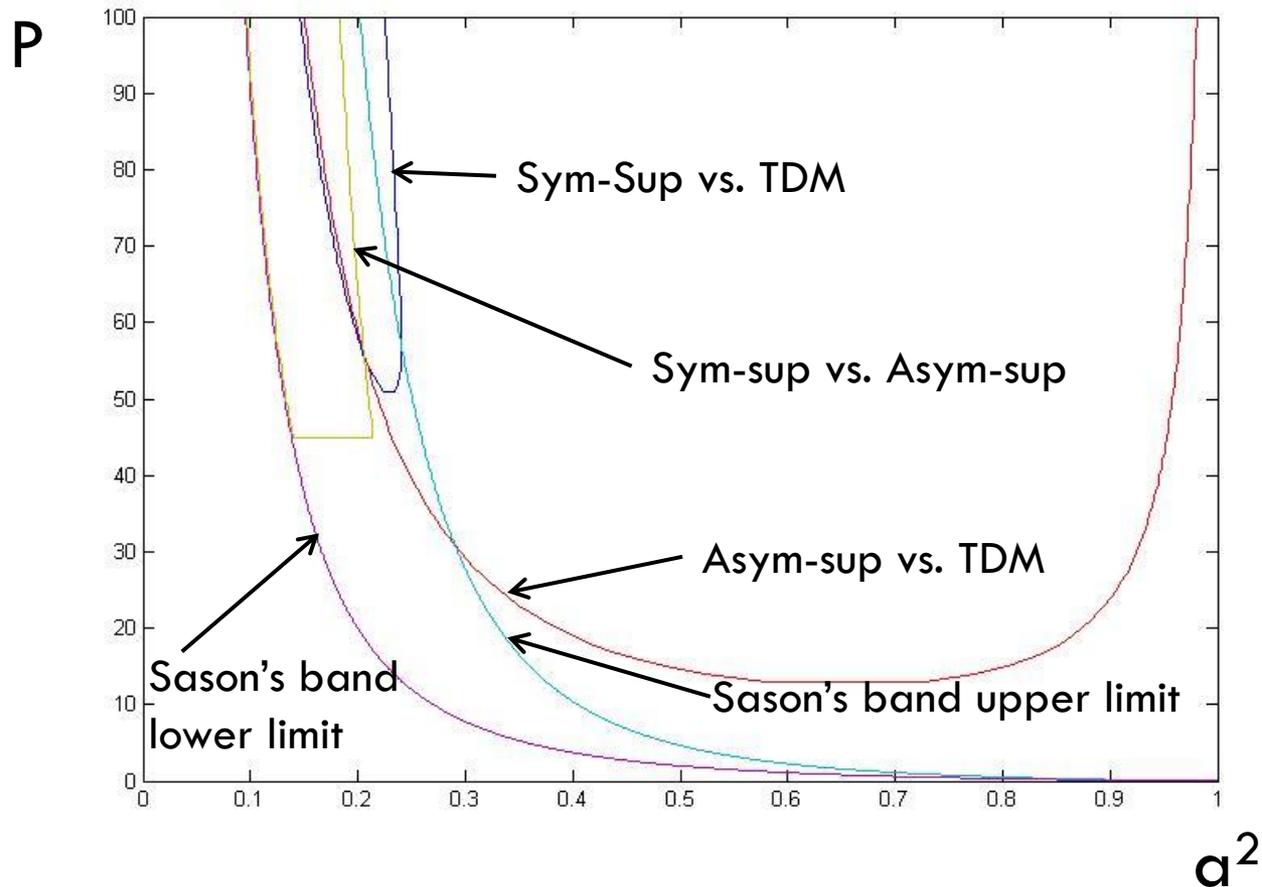


Phase Transitions in Weak Interference

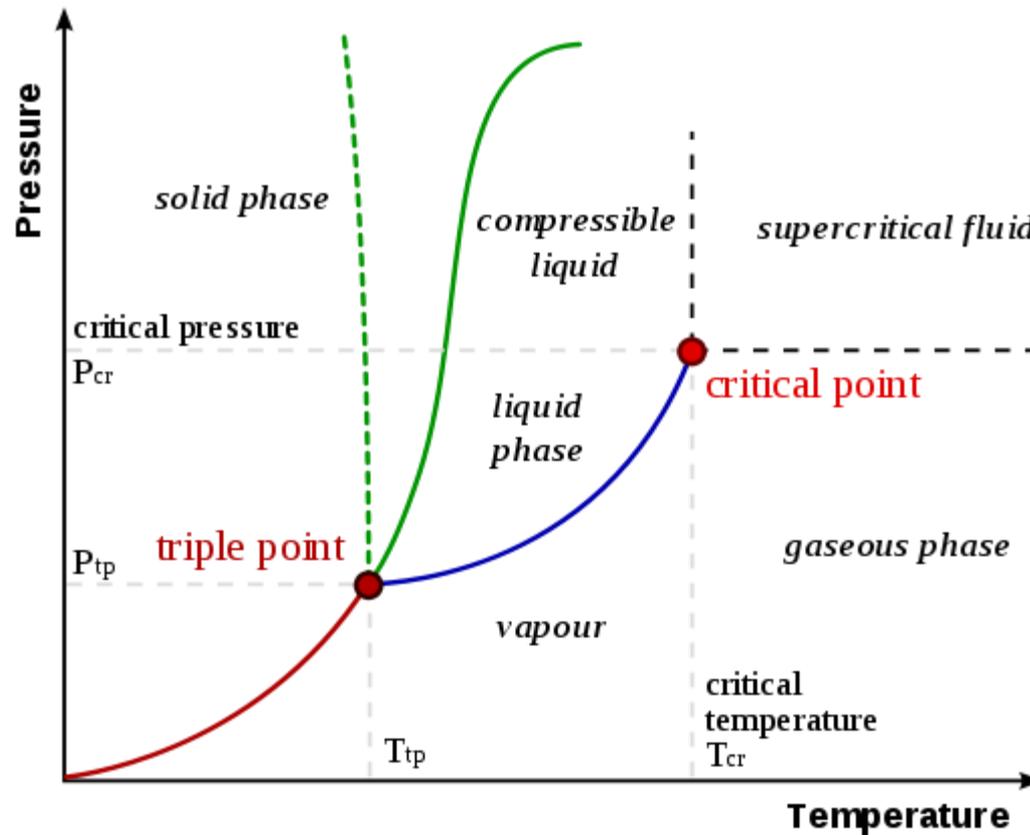
Note: Transitional regions due to convexification along P not included.



Pairwise Phase Transitions



A pleasant resemblance

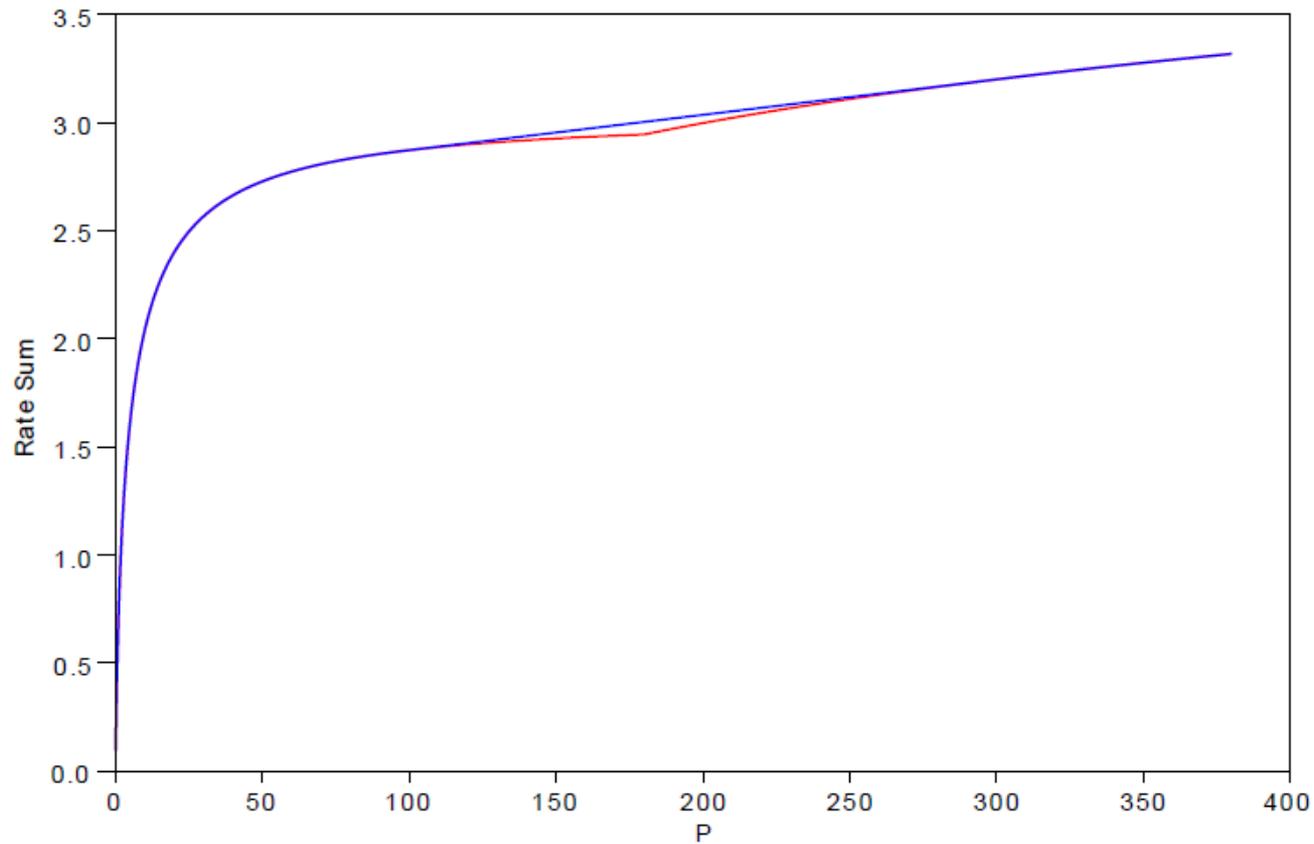


Asymptotically as $P \rightarrow \infty$

$0 < \alpha^2 < 0.087$ -- symmetric superposition is best

$0.087 < \alpha^2 < 1$ – asymmetric superposition is best

As before: Need convexification along P



Final remarks

- Powerful tool: Concave envelopes to transition from one mode to another: time sharing between modes
- Shown a full taxonomy of phase transitions in (a^2, P) parameter space with $0 < a^2 < 1, P > 0$:
- 4 pure modes (IAN, TDM, Symmetric Superposition, and Asymmetric Superposition) and
- 4 transitional regions (IAN vs. TDM, TDM vs. Sym-Sup, TDM vs. Asym-Sup, and Sym-Sup vs. Asym-Sup)

Final remarks

- Working to show this is Gaussian Han-Kobayashi Region
- Future directions: Full achievable region
- Show Gaussian signaling is best
- Converse (capacity region)
- Look at general parameter space (P_1, P_2, a, b)



□ **Many thanks!**